NASA CK-172,140

NASA Contractor Report 172140

ICASE

NASA-CR-172140 19830020654

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Contract No. NAS1-15810 June 1983

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ON THE CONTROL CANONICAL STRUCTURE OF A

CLASS OF SCALAR INPUT SYSTEMS

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ABSTRACT

In this paper, we demonstrate the equivalence of a scalar input system $\dot{x} = \mathcal{A}x + bu$, for which the eigenvalues of the generator \mathcal{A} coincide with the roots of the entire function

$$p(\omega) = e^{\omega T} + a_1 e^{\omega (T-\theta_1)} + \cdots + a_m + \int_0^T a(\theta) e^{\omega (T-\theta)} d\theta,$$

with the controlled scalar functional equation

$$y(t) + a_1 y(t-\theta_1) + \cdots + a_m y(t-T) + \int_0^T a(\theta)y(t-\theta)d\theta = u(t).$$

The theory of nonharmonic Fourier series is then employed to investigate the placement of eigenvalues in the closed-loop system with continuous state feedback.

This research was sponsored in part by the U. S. Air Force Office of Scientific Research under Grant No. AFOSR-79-0018 and in part by the National Aeronautics and Space Administration under NASA Contract No. NASI-15810 while the author was in residence at ICASE, NASA Langley Research Center, Hampton, VA 23665.

1. Introduction

Let $\mathscr A$ be a generator of a strongly continuous group of operators $\{\mathscr G(t): t\in R\}$ on a Hilbert space $\mathscr A$, and suppose that the spectrum of $\mathscr A$ consists of an infinite sequence of simple eigenvalues $\{\omega_k^{}\}$ which forms the zero set of an entire function having the form

$$p(\omega) = e^{\omega T} + a_1 e^{\omega (T-\theta_1)} + \cdots + a_m + \int_0^T a(\theta) e^{\omega (T-\theta)} d\theta.$$
 (1)

In this paper, we will analyze in detail the transformation which carries the scalar input system

$$x(t) = \mathcal{A}x(t) + bu(t), \qquad (2)$$

to the scalar functional equation

$$y(t) + a_1 y(t-\theta_1) + \cdots + a_m y(t-T) + \int_0^T a(\theta)y(t-\theta)d\theta = u(t).$$
 (3)

The latter constitutes the causal control canonical form for the pair (\mathcal{A},b) .

The adjective "causal" is used here because a feedback law of the form

$$u(t) = (f, x(t)) \tag{4}$$

will be shown to transform to a feedback law of the form

$$u(t) = \int_{0}^{T} g(\theta)y(t-\theta)d\theta, \qquad (5)$$

i.e., the input u depends only upon past values of y. The expression

"control canonical form" refers to the fact that the above-mentioned transformation has a structure quite similar to its well-known finite dimensional counterpart as well as to the fact that the effect of linear feedback (4) in the original system (2) can be readily analyzed in terms of the system (3) with corresponding feedback (5).

Before proceeding, we present an example of a system which possesses the structure indicated above. Consider the linear hyperbolic system in characteristic normal form [3]:

$$\frac{\partial}{\partial t} \begin{bmatrix} w \\ w \end{bmatrix} = \begin{bmatrix} \Lambda^{-} & 0 \\ 0 & \Lambda^{+} \end{bmatrix} \frac{\partial}{\partial x} \begin{bmatrix} w \\ w \end{bmatrix} + A(x) \begin{bmatrix} w \\ w \end{bmatrix},$$

with boundary conditions

$$w^{-}(0,t) = D_0 w^{+}(0,t)$$
; $w^{+}(1,t) = D_1 w^{-}(1,t)$.

Here,

$$\Lambda^{-}(x) = diag(\lambda_{1}(x), \dots, \lambda_{n}(x)),$$

$$\Lambda^{+}(x) = diag(\lambda_{n+1}(x), \dots, \lambda_{2n}(x)),$$

are diagonal n×n matrices with

$$\lambda_1(\mathbf{x}) \leq \cdots \leq \lambda_n(\mathbf{x}) < 0 < \lambda_{n+1}(\mathbf{x}) \leq \cdots \leq \lambda_{2n}(\mathbf{x}).$$

Also,

$$A(x) = \begin{bmatrix} A^{-}(x) & A_{12}(x) \\ A_{21}(x) & A^{+}(x) \end{bmatrix},$$

and D_0 and D_1 are both $n \times n$ matrices which determine the manner in which "information" is reflected at the boundaries x = 0 and x = 1. Such systems arise in the study of counterflow heat exchangers [9] and in the study of multiconductor transmission lines [6].

If $A_{12} = A_{21} = 0$ and the off-diagonal components of both A^- and A^+ vanish identically, then the characteristic frequencies are the roots of

$$\pi(\omega) = \det[\exp(\omega\Lambda_{+} + M_{+}) - D_{1}\exp(\omega\Lambda_{-} + M_{-})D_{0}],$$

where

$$\Lambda_{\pm} \equiv \int_{0}^{1} (\Lambda^{\pm}(x)^{-1} dx,$$

and

$$M_{\pm} = \int_{0}^{1} (\Lambda_{\pm}(x)^{-1} A^{\pm}(x) dx.$$

The roots of π will coincide with those of a function p having the form displayed in (1):

$$\pi(\omega) = e^{\omega \alpha} \left(e^{\omega T} + a_1 e^{\omega (T - \theta_1)} + \cdots + a_m \right)$$

$$\equiv e^{\omega \alpha} \cdot p(\omega).$$

In the case where A has non-zero off-diagonal components, one can show [19], under the assumption that both D_0 and D_1 are invertible, that p remains the same except for the addition of an integral term as in (1) with $a(\cdot) \in L^2(0,T)$.

In the above example, one may consider control problems wherein one introduces a forcing term proportional to a scalar function u(t) in the

differential equations (distributed control):

$$\frac{\partial w}{\partial t} = \Lambda(x) \frac{\partial w}{\partial x} + \Lambda(x)w + b(x)u(t), \qquad (6)$$

or in the boundary conditions (point control): e.g.,

$$w^{+}(1,T) = D_{1}w^{-}(1,T) + bu(t).$$
 (7)

Either way, one is led to consider a system of the form

$$\dot{x}(t) = \mathcal{A}x(t) + bu(t),$$

where the element b, in the former case, lies in the state space \mathcal{M} but, in the latter case, must be interpreted as a distribution. We will discuss this point in greater detail in Section 2.

Of central interest in this paper is the spectral synthesis problem: given a set of "desired" eigenvalues $\{\nu_k\}$, can one construct a feedback law of the form (4) such that the eigenvalues of the closed-loop system

$$\dot{\mathbf{x}} = \mathcal{A}\mathbf{x} + b(\mathbf{x}, \mathbf{x}) \equiv (\mathcal{A} + b \otimes \mathbf{x}, \mathbf{x})$$

coincide precisely with $\{\nu_k\}$? Russell [14] has carried out a study of this question in the case corresponding to m=1 above for a class of linear hyperbolic systems consisting of a pair of equations. In his study, the theory of nonharmonic Fourier series is used to study the canonical equation (3) and conclude that any sequence $\{\nu_k\}$ for which

$$\{(\omega_k - \nu_k)/b_k\} \in \ell^2, \tag{8}$$

where $\{b_k\}$ is the sequence of expansion coefficients of the control distribution element b with respect to the eigenfunctions of \mathscr{A} , may be synthesized by continuous linear state feedback. An alternative approach, which apparently avoids an appeal to the theory of nonharmonic Fourier series, has been offered by Clarke and Williamson [2]. Perturbation theorems for spectral operators have been employed by Sun [18] to show that condition (8) is both necessary and sufficient for a class of systems that would include the one studied in this paper. The main contributions of this paper include a more explicit representation of the transformation carrying (2) into (3) than that given in [14] as well as a formula for the expansion coefficients of in terms of the desired set of eigenvalues $\{v_k\}$.

1. A Discrete Finite Dimensional System

For the sake of motivation, we will review briefly the reduction of the finite dimensional discrete system, where A ϵ $\mathbb{R}^{n \times n}$ and b $\epsilon \mathbb{R}^n$,

$$x_k = Ax_{k-1} + bu_k,$$
 (1.1)

to its causal canonical scalar equation

$$y_k + a_1 y_{k-1} + \cdots + a_n y_{k-n} = u_k,$$
 (1.2)

where $det(\lambda I - A) \equiv \lambda^n + a_1 \lambda^{n-1} + \cdots + a_n$

For $r = 1, \dots, n$, we have

$$x_{k+r} = A^{r}x_{k} + \sum_{\ell=1}^{r} A^{r-\ell} bu_{k+\ell}. \qquad (1.3)$$

In particular, if $x_0 = 0$,

$$x_{n} = \begin{bmatrix} A^{n-1}b & \cdots & Ab & b \end{bmatrix} \begin{bmatrix} u_{1} \\ \vdots \\ u_{n} \end{bmatrix} \equiv Cu. \tag{1.4}$$

Well known is the fact that the pair (A,b) is controllable just in case the $n \times n$ matrix C is invertible.

With $a_0 \equiv 1$, we find

$$x_{k+n} + a_1 x_{k+n-1} + \cdots + a_n x_k = \sum_{m=0}^{n-1} a_m x_{k+n-m} + a_n x_k$$

$$= \sum_{m=0}^{n-1} a_m (A^{n-m} x_k + \sum_{\ell=1}^{n-m} A^{n-m-\ell} bu_{k+\ell}) + a_n x_k$$

$$= p(A) x_k + \sum_{\ell=1}^{n} \sum_{m=0}^{n-\ell} a_m A^{n-m-\ell} bu_{k+\ell}, \qquad (1.5)$$

where $p(A) = A^n + a_1 A^{n-1} + \cdots + a_n I = 0$ by the Cayley-Hamilton theorem and where we have interchanged the order of summation in obtaining the last line. The remaining double sum may be rewritten as

$$\sum_{\ell=1}^{n} \sum_{m=0}^{n-\ell} a_m A^{n-m-\ell} b u_{k+\ell} = CM \begin{bmatrix} u_{k+1} \\ \vdots \\ u_{k+n} \end{bmatrix},$$

where

$$M = \begin{bmatrix} 1 & & & \\ a_1 & & & \\ \vdots & & & \\ a_{n-1} & \cdots & a_1 & 1 \end{bmatrix}$$
 (1.6)

Equation (1.5) now reads

$$x_{k+n} + a_1 x_{k+n-1} + \cdots + a_n x_k = CM$$

$$u_{k+n}$$

$$u_{k+n}$$
(1.7)

It is important to note that the coefficients appearing on the left hand side of (1.7) are scalars. Assuming the pair (A,b) to be controllable, or equivalently, the matrix C to be invertible, we define the variable n by

$$\begin{bmatrix} \eta_{k,1} \\ \vdots \\ \eta_{k,n} \end{bmatrix} = M^{-1}C^{-1}x_{k+n} . \tag{1.8}$$

It is also important that $\eta_{k,r}$ depends only on k+r. This may be demonstrated as follows. Let e_r denote the r^{th} standard unit vector in \mathbf{R}^n ; thus

$$\eta_{k,r} = e_r^T M^{-1} C^{-1} x_{k+n}$$

According to (1.8),

$$\begin{bmatrix} n_{k+p,1} \\ \vdots \\ n_{k+p,n} \end{bmatrix} = M^{-1}C^{-1}x_{k+p+n} .$$

Suppose p+q = r. We wish to show that

$$n_{k,r} = n_{k+p,q}$$
,

i.e, that

$$e_r^{T_M-1}c^{-1}x_{k+n} = e_q^{T_M-1}c^{-1}x_{k+p+n}$$
.

Since

$$x_{k+p+n} = A^{p}x_{k+n} + \sum_{\ell=1}^{p} A^{p-\ell} bu_{k+\ell+n},$$

we must have:

(i)
$$e_r^T = e_q^T M^{-1} C^{-1} A^p C M = e_q^T \hat{A}^p$$

(ii)
$$e_q^T M^{-1} C^{-1} \sum_{\ell=1}^{p} A^{p-\ell} bu_{k+n-\ell} = 0.$$

That (i) is true follows from the well-known fact (see [10] or [16], e.g.) that \hat{A} is the companion matrix of A:

$$\hat{A} = M^{-1}C^{-1}ACM = \begin{bmatrix} 0 & 1 & & & \\ 0 & 0 & 1 & & \\ \vdots & & & & \\ -a_n & & \cdots & -a_2 & -a_1 \end{bmatrix}.$$

That (ii) holds true follows from the observation that

$$C^{-1}A^{p-\ell}b = e_{n-p+\ell}$$
,

and the fact that $\operatorname{span}\{e_{n-p+\ell},\cdots,e_n\}$ is invariant under M^{-1} .

We now define the sequence $\{y_k\} \subset \mathbb{R}$ by

$$y_{k+1} = n_{k,1}$$

Equation (1.8) now reads

$$\begin{bmatrix} y_{k+1} \\ \vdots \\ y_{k+n} \end{bmatrix} = M^{-1}C^{-1}x_{k+n}.$$

Multiplying (1.7) on the left by $M^{-1}C^{-1}$, we find that $\{y_k\}$ satisfies (1.2).

Let us now see how linear feedback in (1.1) manifests itself in (1.2).

Ιf

$$u_k = f^T x_{k-1}$$
, $f^T = row vector$,

then, by 1.8,

$$u_{k} = f^{T}CM \begin{bmatrix} y_{k-n} \\ \vdots \\ y_{k-1} \end{bmatrix} \equiv g^{T} \begin{bmatrix} y_{k-n} \\ \vdots \\ y_{k-1} \end{bmatrix}, \qquad (1.9)$$

where $g^T \equiv (g_n, \dots, g_1)$. Thus, the closed-loop system

$$x_k = (A + bf^T)x_{k-1}$$
,

goes over to the closed-loop system

$$y_k + (a_1 - g_1)y_{k-1} + \cdots + (a_n - g_n)y_{k-n} = 0.$$

Let $\lambda_1, \dots, \lambda_n$ denote the (not necessarily distinct) eigenvalues of A. We are assuming that the entries in A are all real; hence $\{\lambda_1, \dots, \lambda_n\}$ is symmetric with respect to the real axis in the complex plane. Given any likewise symmetric set of complex numbers $\{\nu_1, \dots, \nu_n\}$, it is now an easy matter to construct a feedback gain $f \in \mathbb{R}^n$ for which the eigenvalues of A + bf^T are precisely $\{\nu_1, \dots, \nu_n\}$. Indeed, if

$$\lambda^{n} + c_{1}\lambda^{n-1} + \cdots + c_{n} = \prod_{i=1}^{n} (\lambda - v_{i}),$$

then $c_i \in \mathbb{R}$ for $i=1,\dots,n$ and hence, referring to (1.10), we must have

$$g_i = a_i - c_i$$
 $i=1,\dots,n$

òr

$$g^{T} = a^{T} - c^{T}$$
 where $a^{T} = (a_{n}, \dots, a_{1})$, etc.

Referring to (1.9), we arrive at the Bass-Gura formula [10] for f:

$$f^{T} = (a-c)^{T}M^{-1}C^{-1}$$
. (1.11)

We will show that the same formula holds true under appropriate assumptions for the class of infinite dimensional systems mentioned above and to which we now turn.

2. Nonharmonic Fourier Series and Controllability

In this section, we discuss the nature of the transformation which corresponds to the matrix C of the previous section. Before doing so, it is first necessary to recall some definitions which concern the scalar input system

$$\dot{x}(t) = \mathcal{A}x(t) + bu(t). \tag{2.1}$$

Let $\mathscr A$ generate a strongly-continuous group $\{\mathscr G(t):t\in\mathbb R\}$ on the Hilbert space $\mathscr A$, and denote the action of a conjugate linear functional x on $\psi\in\mathscr A$ by $\langle\psi,x\rangle$. As with the inner product (\cdot,\cdot) on $\mathscr A$, the bracket $\langle\cdot,\cdot\rangle$ is understood to be conjugate linear in the first argument and linear in the second. It is also understood that $\langle\cdot,\cdot\rangle$ has the property that $\langle\psi,x\rangle=(\psi,x)$ if $x\in\mathscr A$. Let us define the weak solution $x(\cdot)$ to (2.1) satisfying $x(0)=x_0\in\mathscr A$ by requiring

$$\langle \psi, x(t) \rangle = \langle \mathscr{G}(t)^* \psi, x_0 \rangle + \int_0^t \langle \mathscr{G}(t-s)^* \psi, b \rangle u(s) ds,$$
 (2.2)

for all $\psi \in \mathcal{D}(\mathring{\mathcal{J}})$. Here, $\{\mathscr{G}(t)^*: t \in \mathbb{R}\}$ denotes the adjoint group generated by \mathscr{A}^* with domain $\mathscr{D}(\mathring{\mathcal{J}}^*) \subset \mathscr{U}$ and b is understood to represent a conjugate linear functional on \mathscr{U} .

Definition 2.1. [4]:

The control distribution element b is said to be admissible with respect to $\mathcal A$ if b can be identified with a conjugate linear functional, also denoted by b, for which

(i)
$$\mathcal{D}(b) \supseteq \mathcal{D}(\mathcal{A}^{\dagger})$$
.

(ii) for any initial state $x(0) \in \mathcal{U}$ and any locally square integrable input u, the weak solution x(t) of (2.1) lies in \mathcal{U} for each t > 0.

We note that if $\checkmark \epsilon \not A$ in (2.1), then the weak solution coincides with the mild solution

$$x(t) = \mathcal{G}(t)x_0 + \int_0^t \mathcal{G}(t-s)bu(s)ds. \qquad (2.3)$$

Since $x(t) \in \mathcal{H}$ for each t > 0 [5], b in this case is admissible.

To make (2.2) somewhat more constructive, we introduce the sequence $\{\phi_k\}$ of eigenfunctions of $\mathscr A$ with corresponding eigenvalues $\{\omega_k\}$. We will assume that each ω_k is simple and that $\{\phi_k\}$ forms a Riesz basis for $\mathscr A$:

Definition 2.2 ([7], [11]):

A sequence $\{\phi_k\}$ of elements of a Hilbert space is said to form a Riesz basis for \mathcal{H} if the exist constants d,D with $0 < d \le D < \infty$ such that every $x \in \mathcal{H}$ may be expanded uniquely in a series

$$x = \sum_{k} x_{k} \phi_{k}$$

with

$$d^{2} \sum_{k} |x_{k}|^{2} \le |x_{k}|^{2} \le D^{2} \sum_{k} |x_{k}|^{2}.$$
 (2.4)

Equivalently ([7], e.g.), a Riesz basis may be thought of as the image of an orthonormal basis $\{e_k\}$ of \mathcal{H} under a bounded and boundedly invertible transformation Φ :

$$\phi_{\mathbf{k}} = \Phi \mathbf{e}_{\mathbf{k}}$$

With every Riesz basis $\{\phi_k\}_{\text{t}}$ there is associated a unique "dual" Riesz basis $\{\psi_{\text{L}}\}$ defined by

$$\psi_{\mathbf{k}} = (\Phi^*)^{-1} \mathbf{e}_{\mathbf{k}},$$

with the biorthonormality property $(\psi_k, \phi_k)_{\mu} = \delta_{k\ell}$. The expansion coefficients of $x \in \mathcal{H}$ with respect to $\{\phi_k\}, \{\psi_k\}$ are $(\psi_k, x), (x, \phi_k)$ respectively.

Let us denote by $\{\psi_k\}$ the sequence of eigenfunctions of \mathscr{A}^* ; i.e., $\mathscr{A}^*\psi_k = \overline{\psi}_k \psi_k$. Suitably normalized, $\{\psi_k\}$ is easily seen to be dual to $\{\phi_k\}$. Now let $\psi = \psi_k$ in (2.2) and set $x_k(t) = \langle \psi_k, x(t) \rangle$ and $b_k = \langle \psi_k, b \rangle$. Then

$$x_k(t) = e^{\omega_k t} x_k(0) + b_k \int_0^t e^{\omega_k (t-s)} u(s) ds,$$
 (2.5)

and hence, weak solutions may be represented as

$$x(t) = \sum_{k} x_k(t) \phi_k$$

with the x_k 's given by (2.5). If $\sup |Re \omega_k| < \infty$, admissibility reduces, via (2.4), to whether or not, for each t > 0, the sequence

$$\{b_k \int_0^t e^{u_k(t-s)} u(s)ds\} \in \ell^2,$$

for every $u \in L^2(0,t)$.

In the course of reducing (2.1) to canonical form, we will carry out manipulations similar to those employed in the previous section and which involve the expression

$$\int_{0}^{T} \mathcal{G}(T-s)bu(s)ds.$$

The whole point of the above discussion is that the latter expression may, even if $b \notin \mathcal{A}$, be interpreted rigorously if the eigenfunctions $\{\phi_k\}$ of form a Riesz basis for \mathcal{A} and b is admissible with respect to \mathscr{A} :

$$\int_{0}^{T} \mathcal{G}(t-s)bu(s)ds = \sum_{k} \left(b_{k} \int_{0}^{t} e^{\omega_{k}(t-s)} u(s)ds\right)\phi_{k}.$$

In particular, if $x_0 = 0$, we may write

$$\mathbf{x}(\mathbf{T}) = \int_{0}^{\mathbf{T}} \mathcal{G}(\mathbf{T} - \mathbf{s}) b\mathbf{u}(\mathbf{s}) d\mathbf{s} \equiv \mathcal{G}(\mathbf{T}) \mathbf{u}. \tag{2.6}$$

This is the analogue of (1.4). Conversely, if $x \in \mathcal{H}$, the problem of constructing a control $u \in L^2(0,T)$ which takes the origin to x at time T is characterized by a moment problem:

$$\mathcal{L}(T)u = x$$

or, with $x = \sum_{k} x_k \phi_k$,

$$b_k \int_0^T e^{\omega_k (T-s)} u(s) ds = x_k.$$
 (2.7)

<u>Definition 2.3</u> ([15]):

The pair (\mathcal{A},b) is said to be approximately, exactly controllable in time T if $\mathcal{C}(T):L^2(0,T)\to\mathcal{H}$ is densely, boundedly invertible respectively.

Referring to (2.7), it is obvious that approximate controllability requires $b_k \neq 0$ for all k. But controllability also depends upon properties of the sequence of exponentials $\{e^{\omega_k}^i\}$. This paper is concerned with a very special class of \mathscr{A} 's, namely those whose spectrum coincides with the zero set $\{\omega_k\}$ of a function having the form

$$p(\omega) = e^{\omega T} + a_1 e^{\omega (T-\theta_1)} + \cdots + a_m + \int_0^T a(\theta) e^{\omega (T-\theta)} d\theta. \qquad (2.8)$$

It is assumed that $a_i \in \mathbb{R}$, $i=1,\dots,m$ with $a_m \neq 0$, that $a(\cdot) \in L^2(0,T)$ is real valued, and that

$$0 < \theta_1 < \cdots < \theta_m \equiv T$$
.

Let us write

$$p(\omega) = p_0(\omega) + \int_0^T a(\theta)e^{\omega(T-\theta)} d\theta,$$

and let $\{\sigma_k^{}\}$ denote the zero set of the exponential polynomial p_o . It is easy to see that the set $\{\sigma_k^{}\}$ as well as the set $\{\omega_k^{}\}$ must lie in a vertical strip of finite width in the complex plane. It has been shown [1] using the argument principle that an infinite number of $\sigma_k^{}$'s do exist and that the number of $\sigma_k^{}$'s in any horizontal strip of fixed height is bounded.

The prototype in this situation is $p_o(\omega) = e^{2\omega} - 1$ which has for its set of zeros $\{\sigma_k\} = \{k\pi i\}$. It can also be shown [17], [19], using the above properties of $\{\sigma_k\}$ together with the Fourier transform in the complex domain, that the sequence of exponentials $\{e^{\alpha}k^{\bullet}\}$ forms a Riesz basis for $L^2(0,T)$ if, in addition, inf $|p_o'(\sigma_k)| > 0$. The latter condition ensures the bounded invertibility of the map Φ discussed after Definition 2.2. One can also demonstrate using Rouche's theorem that the sequence $\{\omega_k\}$ of zeros of p may be indexed in such a way that $\{\omega_k - \sigma_k\} \in \ell^2$ and that

inf $|p'(\omega_k)| > 0$ is sufficient to yield the Riesz basis property for the k sequence $\{e^{k}\}$. We omit the somewhat detailed proof of these statements [19] and return to their relationship with the notion of controllability.

Let us denote the sequence biorthonormal to $\{e^{\omega_k}\}$ by $\{q_{\mathfrak{q}}(\cdot)\}$:

$$\int_{0}^{T} \frac{1}{q_{\ell}(s)} e^{\omega_{k}^{S}} ds = \delta_{k\ell}.$$

The construction of the sequence $\{q_{\ell}(\cdot)\}$ goes back to the work of Paley and Wiener [12] and is carried out as follows. One forms the function

$$\hat{q}(\omega) = \frac{p(\omega)}{p'(\omega_{L})(\omega - \omega_{L})}$$

and notes that $\hat{q}_{\ell}(\omega_{k}) = \delta_{k\ell}$. The Paley-Wiener theorem ([12],[13]) then asserts that \hat{q}_{ℓ} is the Laplace transform of a square integrable function $\overline{q}_{\ell}(\cdot)$ whose support $\sup(q_{\ell}) \subseteq [0,T]$:

$$\hat{q}_{\ell}(\omega) = \int_{0}^{T} e^{\omega s} \overline{q}_{\ell}(s) ds.$$

Thus, \overline{q}_{ℓ} may be represented via the inverse Laplace transform

$$\overline{q}_{\ell}(s) = \frac{1}{2\pi i} \int_{-i\infty}^{i\infty} e^{-s\omega} \hat{q}_{\ell}(\omega) d\omega. \qquad (2.9)$$

The sequence $\{q_{\ell}(\cdot)\}$ so constructed is biorthonormal to $\{e^{u_{k}^{\dagger}}\}$ since

$$\int_0^T \overline{q}_{\ell}(s) e^{\omega_k s} ds = \hat{q}_{\ell}(\omega_k) = \delta_{k\ell}.$$

However, in the next section, we shall derive an explicit formula for each q_{ℓ} without any mention of the Paley-Wiener theory. In effect, the inverse transform (2.9) can be computed explicitly in closed form.

Assuming $b_k \neq 0$ for any k, we may now invert relationship (2.7) with the aid of the sequence $\{q_k(\bullet)\}$. Indeed, the control

$$u(t) = \sum_{k} \frac{x_k}{b_k} \overline{q}_k (T-t),$$

satisfies (2.7). Here, T is understood to be the quantity appearing in the definition (2.8) of p. Thus $\mathcal{L} = \mathcal{L}(T)$ has the properties

$$\mathscr{L}: \ \overline{q}_{k}(T^{-\bullet}) \rightarrow b_{k} \phi_{k} \tag{2.10}$$

$$\mathscr{L}^{-1}: \phi_k + b_k^{-1} \overline{q}_k (T^{-\bullet}),$$
 (2.11)

which will be used throughout the remainder of the paper.

We conclude this section with a summary of the various assumptions made and a brief discussion of their significance.

Assumption 2.4:

- (i) $\mathcal A$ generates a strongly-continuous group $\{\mathscr G(t):t\in R\}$ on a Hilbert space $\mathcal H$.
- (ii) The sequence of eigenfunctions $\{\phi_k\}$ of $\mathscr A$ forms a Riesz basis for $\mathscr H$.
- (iii) The eigenvalues $\{\omega_k^{}\}$ of $\mathscr A$ are the roots of $p(\omega)$ given by (2.8) and $p'(\omega_k^{}) \neq 0$ for any k.
 - (iv) The control distribution element b is admissible with respect to with coefficients $b_k = \langle \psi_k, b \rangle \neq 0$ for any k. $(\bigcirc_k^* \psi_k = \overline{\omega}_k \psi_k)$.

Since finite linear combinations of the ϕ_k 's are dense in \mathcal{H} , the pair (\mathcal{A},b) , under the above assumption, is approximately controllable in time T (c.f. [15]). If $\inf_k |b_k| > 0$ and $\inf_k |p'(\omega_k)| > 0$, exact controllability in time T may be established as well [15].

Part (iii) of the above assumption essentially limits the scope of our results to the class of linear hyperbolic systems described in the introduction. For such systems, $p'(\omega_k) \neq 0$ for all k does not hold true in general. We make this hypothesis here only to simplify our presentation — notwithstanding a good dose of tedium, it can be shown by introducing generalized exponentials (i.e., $t^{j_e}^{\omega_k t}$) in the case of multiple roots ω_k that the canonical form presented in this paper remains valid.

The case of boundary control (see equation (7)), where $b \in \mathbb{R}^n$, does correspond to an admissible distribution b ([15], Theorem 3.1)

$$b_k = \langle \psi_k, b \rangle = \psi_k^+(1)^* \Lambda^+(1)b,$$

and, if no component of b vanishes, approximate countrollability may be established [19]. We do not wish to emphasize the abstract character of the above assumptions. However, the development of a canonical form is carried out most transparently in this setting.

3. Reduction to Canonical Form

In this section, we will derive the transformation which carries the system (2) to its canonical form (3), and in order to pursue as closely as possible the analogy between the pair (A,b) of the discrete system above with the pair (A,b) at hand, we will proceed somewhat formally. Assumption

2.4 will be understood to be in force throughout. We begin with an analogue of the Cayley-Hamilton theory:

Lemma 3.1: Under (i), (ii), and (iii) of Assumption 2.4,

$$\mathcal{G}(t) + a \mathcal{G}(t-\theta_1) + \cdots + a \mathcal{G}(t-T) + \int_0^T a(\theta) \mathcal{G}(t-\theta) d\theta = 0.$$

<u>Proof.</u> Given an arbitrary initial state $x = \sum_{k} x_k \phi_k \in \mathcal{H}$, the solution to $\dot{x}(t) = \mathcal{A}x(t)$ is given formally by $x(t) = \mathcal{G}(t)x$ and concretely by

$$x(t) = \sum_{k} e^{\omega_k t} x_k \phi_k$$

Thus

$$\begin{aligned} \left[\mathcal{G}(t) + a_1 \mathcal{G}(t - \theta_1) + \cdots + a_m \mathcal{G}(t - T) + \int_0^T a(\theta) \mathcal{G}(t - \theta) d\theta \right] x \\ &= \sum_k p(\omega_k) e^{\omega_k (t - T)} \\ x_k \phi_k = 0. \end{aligned}$$

This completes the proof of the lemma.

Formally, solutions to $\dot{x}(t) = \mathcal{A}x(t) + bu(t)$ satisfy

$$x(t+r) = \mathcal{G}(r)x(t) + \int_{0}^{r} \mathcal{G}(r-s)bu(s)ds.$$
 (3.1)

Thus,

$$x(t+T) + a_{1}x(t+T-\theta_{1}) + \cdots + a_{m}x(t) + \int_{0}^{T} a(\theta)x(t+T-\theta)d\theta$$

$$= \left[\mathcal{G}(t) + a_{1}\mathcal{G}(t-\theta_{1}) + \cdots + a_{m}\mathcal{G}(t-T) + \int_{0}^{T} a(\theta)\mathcal{G}(t-\theta)d\theta\right]x(T)$$

$$+ \int_{0}^{T} (T-s)bu(t+s)ds + a_{1}\int_{0}^{T-\theta_{1}} \mathcal{G}(T-\theta_{1}-s)bu(t+s)ds$$

$$+ \cdots + a_{m-1}\int_{0}^{T-\theta_{m-1}} \mathcal{G}(T-\theta_{m-1}-s)bu(t+s)ds$$

$$+ \int_{0}^{T} a(\theta) \int_{0}^{T-\theta} \mathcal{G}(T-\theta-s)bu(t+s)ds d\theta. \qquad (3.2)$$

The first term is zero by Lemma 3.1. Changing the inner variable of integration from s to $s+\theta$ and then interchanging the order of integration, the last term may be written as

$$\int_{0}^{T} \mathcal{G}(T-s)b \int_{0}^{s} a(\theta)u(t+s-\theta)d\theta ds.$$

Likewise,

$$\int_{0}^{T-\theta_{i}} \mathcal{G}(T-\theta_{i}-s)bu(t+s)ds = \int_{0}^{T} \mathcal{G}(T-\theta_{i}-s)b\chi_{[0,T-\theta_{i}]}(s)u(t+s)ds$$

$$= \int_{0}^{T} \mathcal{G}(T-s)b\chi_{[0,T-\theta_{i}]}(s-\theta_{i})u(t+s-\theta_{i})ds$$

$$= \int_{0}^{T} \mathcal{G}(T-s)b\chi_{[\theta_{i},T]}(s)u(t+s-\theta_{i})ds,$$

where $\chi_{[a,b]}(\cdot)$ is the characteristic function of the interval [a,b]. The right-hand side of (3.2) may now be written as

$$\int_{0}^{T} \mathcal{G}(T-s)b(\mathcal{M}u(t+\bullet))(s)ds = \mathcal{C}\mathcal{M}u(t+\bullet),$$

where $\mathscr L$ is as before and $\mathscr M\colon L^2(0,T)\to L^2(0,T)$ is the transformation defined by

$$(My)(s) = y(s) + a_1 \chi_{[\theta_1,T]}(s)y(s-\theta_1) + \cdots$$

$$+ a_{m-1} \chi_{[\theta_{m-1},T]}(s)y(s-\theta_{m-1}) + \int_0^s a(\theta)y(s-\theta)d\theta. \tag{3.3}$$

Equation (3.2) now reads

$$x(t+T) + a_1 x(t+T-\theta_1) + \cdots + a_m x(t) + \int_0^T a(\theta) x(t+T-\theta) d\theta = \mathcal{L} ku(t+\cdot).$$
 (3.4)

Equations (3.3) and (3.4) should be compared with (1.6) and (1.7) respectively. To proceed further, the nature of the transformation $\mathcal M$ must be analyzed.

Lemma 3.2:

- (i) \mathcal{M} is bounded and boundedly invertible.
- (ii) If $g \in L^2(0,T)$ with $supp(g) \subset [T-\theta,T]$ for some $\theta \in (0,T)$, then $supp(\mathcal{M}^{-1}g) \subset [T-\theta,T]$ as well.

Proof: That \mathcal{M} is bounded is clear from (3.3). Referring to the latter, let us write

$$\mathcal{M} = \mathcal{I} - \mathcal{N}_1 - \mathcal{N}_2,$$

with $(\mathcal{N}_2 y)(s) \equiv -\int_0^s a(\theta)y(s-\theta)d\theta$. A standard theorem from integral

equations [8] yields the fact that $(\mathcal{P}-\mathcal{N}_2)^{-1}$ exists and is bounded with

$$(\mathcal{I}-\mathcal{N}_2)^{-1} = \mathcal{I} + \sum_{k=1}^{\infty} \mathcal{N}_2^k$$
.

It is easily established that $\sup(\mathcal{N}_2 g)$ and hence $(\mathcal{F}-\mathcal{N}_2)^{-1}g) \subseteq [T-\theta,T]$ if $\sup(g) \subseteq [T-\theta,T]$. Moreover, \mathcal{N}_1 is nilpotent since $\sup(\mathcal{N},g) \subseteq [T-\theta+\theta_1,T]$ if $\sup(g) \subseteq [T-\theta,T]$. Thus $(\mathcal{F}-\mathcal{N}_2)^{-1}\mathcal{N}_1$ is nilpotent and

$$\mathcal{M}^{-1} = (\mathcal{I} - \mathcal{N}_2)^{-1} (\mathcal{I} - (\mathcal{I} - \mathcal{N}_2)^{-1} \mathcal{N}_1)^{-1}$$

$$= (\mathcal{I} - \mathcal{N}_2)^{-1} (\mathcal{I} + \sum_{k=1}^{n} ((\mathcal{I} - \mathcal{N}_2)^{-1} \mathcal{N}_1)^k),$$

for some $n < \infty$, which shows that \mathcal{M}^{-1} is bounded and satisfies (ii). This completes the proof.

We remark that part (ii) of Lemma 3.2 is the analogue of the fact that $span\{e_{n-p}, \cdots, e_n\}$ is invariant under the matrix M^{-1} discussed in Section 1. We next Fourier analyze the action of $\mathcal M$ on the exponentials e^{ω_k} :

Lemma 3.3:
$$\int_{0}^{T} e^{\omega(t-s)} (e^{\omega k})(s) ds = p(\omega)/(\omega - \omega_{k}).$$

Proof: In forming the expression on the left, we find terms of the form

$$\int_{0}^{T} e^{\omega(T-s)} \chi_{[\theta_{i},T]} (s) e^{\omega_{k}(s-\theta_{i})} ds,$$

and the term

$$\int_{0}^{T} e^{u_{k}(T-s)} \int_{0}^{s} a(\theta)e^{u_{k}(s-\theta)} d\theta ds.$$

Standard manipulations reduce the former to

$$\frac{e^{\omega(T-\theta_1)}-e^{\omega_k(T-\theta_1)}}{\omega-\omega_k},$$

and the latter to

$$\int_{0}^{T} a(\theta) \frac{e^{\omega(T-\theta)} - e^{\omega_{k}(T-\theta)}}{\omega - \omega_{k}} d\theta.$$

It follows that

$$\int_{0}^{T} e^{\omega(T-s)} \mathscr{M} e^{k} (s) ds = \left[e^{\omega T} - e^{k} + a_{1} \left(e^{\omega(T-\theta_{1})} - e^{k} (T-\theta_{1}) \right) + \cdots + a_{m-1} \left(e^{\omega(T-\theta_{m-1})} - e^{\omega_{k} (T-\theta_{m-1})} \right) + \int_{0}^{T} a(\theta) \left(e^{\omega(T-\theta)} - e^{\omega_{k} (T-\theta)} \right) d\theta \right] / (\omega - \omega_{k})$$

$$= \left(p(\omega) - p(\omega_{k}) \right) / (\omega - \omega_{k}) = p(\omega) / (\omega - \omega_{k}),$$

since $\boldsymbol{\omega}_k$ is a zero of $\ p.$ This completes the proof.

Recall that the basis $\{e^{\omega_k}\} \subset L^2(0,T)$ has a unique biorthonormal sequence which we denote by $\{q_k(\cdot)\} \subset L^2(0,T)$. The key property which $\mathcal M$ possesses is the following:

Lemma 3.4:
$$\mathscr{M}: e^{\omega_k^{\bullet}} \rightarrow p'(\omega_k)\overline{q_k}(T^{-\bullet}).$$

Proof: According to Lemma 3.3,

$$\int_{0}^{T} e^{\omega_{j}(T-s)} (e^{\omega_{k}})(s) ds = p'(\omega_{k}) \delta_{jk}.$$

Since the sequence $\{q_j(T^{-\bullet})\}$ is biorthonormal to $\{e^{ij}\}$ and unique, we conclude that

$$\mathcal{M}e^{\omega_{k}^{\bullet}})(s) = p'(\omega_{k})\overline{q}_{k}(T-s),$$

which completes the proof.

Remark:

The preceding lemma allows us to construct the biorthonormal sequence explicitly:

$$\overline{q}_k(s) = \frac{1}{p'(\omega_k)} \mathcal{M}e^{\omega_k^{\bullet}})(T-s).$$

Under Assumption 2.4, (iv), the map $\mathscr L$ possesses a densely defined inverse $\mathscr L^{-1}$ (see (2.11)). If $x(\cdot)$ is a solution of $x = \mathscr A x + bu$ with sufficiently regular initial state x_0 and input $u(\cdot)$, we may define

$$\eta(t, \bullet) = \mathcal{M}^{-1} \mathcal{L}^{-1} x(t+T) (\bullet) \in L^{2}(0,T).$$
(3.5)

Let us assume that n is defined for every t and s. Then

Lemma 3.5: n(t,s) is a function of s+t.

<u>Proof:</u> Suppose p+s=r < T. We wish to show that n(t,r)=n(t+p,s). According to (3.5),

$$\eta(t,r) = \mathcal{U}^{-1} \mathcal{L}^{-1} x(t+T)(r).$$

If we let $x(t+T) = \sum_{k} x_k(t+T)\phi_k$ and use (2.11) and Lemma 3.4, we have

$$\eta(t,r) = \sum_{k} p'(\omega_{k})^{-1} b_{k}^{-1} x_{k}(t+T) e^{\omega_{k} r}$$

Likewise,

$$\eta(t+p,s) = \sum_{k} p'(\omega_{k})^{-1} b_{k}^{-1} x_{k}(t+p+T) e^{\omega_{k} s}$$
.

Since

$$x_{k}(t+p+T) = e^{u_{k}p} x_{k}(t+T) + b_{k} \int_{0}^{p} e^{u_{k}(p-\theta)} u(t+\theta+T)d\theta,$$

we have

$$\eta(t+p,s) - \eta(t,r) = \sum_{k} (p'(\omega_{k})^{-1})^{p} e^{\omega_{k}(p-\theta)} u(t+\theta+T)d\theta e^{\omega_{k}s}.$$

If we change the variable of integration via $p-\theta=T-\tau$ and invoke Lemma 3.4, the right-hand side becomes

$$= \mathcal{M}^{-1} \sum_{k}^{T} \int_{T-p}^{w_k(T-\tau)} e^{u(t+p+\tau)d\tau} \overline{q}_k^{(T-\bullet)}(s).$$

But the function on which \mathscr{M}^{-1} operates is nothing but $\chi_{[T-p,T]}(\cdot)u(t+p+\cdot)$ expanded in terms of the sequence $\{\overline{q}_k(T-\cdot)\}$. Hence

$$= \left(\mathcal{N}^{-1} \chi_{[T-p,T]}(\cdot) u(t+p+\cdot) \right)(s)$$

$$\equiv \left(\mathcal{N}^{-1} g \right)(s).$$

Clearly, supp(g) \subset [T-p,T], and since $0 \le p+s = r \le T$ implies $s \in [0,T-p]$,

we conclude from Lemma 3.2 that $\mathscr{U}^{-1}g)(s) = 0$. Thus, $\eta(t+p,s) = \eta(t,r)$. This completes the proof.

Lemma 3.5 may be shown to hold true in the case where $\eta(t, \bullet)$ is merely a square-integrable function for each t by approximating everything with continuous functions and then passing to the limit. We may now define the scalar variable

$$y(t+\cdot) = \eta(t,\cdot) = ((-1)^{-1}x(t+T))(\cdot).$$
 (3.6)

We conclude this section with

Theorem 3.6: Under Assumption (2.4), the variable y defined by (3.6) satisfies the functional equation

$$y(t) + a_1 y(t-\theta_1) + \cdots + a_m y(t-T) + \int_0^T a(\theta) y(t-\theta) d\theta = u(t).$$
 (3.7)

Proof: Apply $\mathcal{M}^{-1}\mathcal{L}^{-1}$ on the left to both sides of (3.4).

4. Spectral Synthesis

Let us first examine how continuous state feedback

$$u(t) = (f, x(t))$$

in the original system (2.1) manifests itself in the canonical system (3.7).

Let 6 have the expansion

$$\delta = \sum_{j} f_{j} \psi_{j},$$

where $\{\psi_j\}$ denotes the sequence of eigenfunctions of \mathscr{A}^* . Let x(t) and $y(t+\cdot)$ have the expansions

$$x(t) = \sum_{k} x_{k}(t) \phi_{k}$$

$$y(t+\bullet) = \sum_{k} y_{k}(t) e^{\omega_{k} \bullet} .$$
(4.1)

Referring to (3.6), we have

$$x_k(t+T) = b_k p'(\omega_k) y_k(t)$$

Thus,

$$u(t) = (\delta, x(t)) = \sum_{k} f_k b_k p'(\omega_k) y_k(t-T).$$

Again denoting the sequence biorthonormal to $\{e^{k}\}$ by $\{q_{j}(\cdot)\}$, the right-hand side of the last equation may be written as

$$\int_{0}^{T} \left(\sum_{j} f_{j} b_{j} p'(\omega_{j}) \overline{q}_{j}(s) \right) \left(\sum_{j} y_{k}(t-T) e^{\omega_{k}^{s}} \right) ds.$$

By (4.1),

$$\sum_{k} y_{k}(t-T)e^{\omega_{k}^{s}} = y(t-T+s).$$

If we change the variable of integration via $s = T - \theta$ and define

$$g(\theta) \equiv \sum_{j} f_{j} b_{j} p'(\omega_{j}) \overline{q}_{j} (T-\theta),$$
 (4.2)

we obtain

$$u(t) = \int_{0}^{T} g(\theta)y(t-\theta)d\theta.$$

The resulting closed-loop canonical system is simply

$$y(t) + a_1 y(t-\theta_1) + \cdots + a_m y(t-T) + \int_0^T (a(\theta)-g(\theta))y(t-\theta)d\theta = 0.$$
 (4.3)

Thus, the eigenvalues of the closed-loop system

$$x(t) = (\mathcal{A} + b \otimes \{ \})x(t)$$

are roots of

$$p_{g}(\omega) = p_{o}(\omega) + \int_{0}^{T} (a(\theta) - g(\theta)) e^{\omega(T-\theta)} d\theta. \qquad (4.4)$$

As we have already mentioned in Section 2, one can show using Rouche's theorem that the sequence $\{v_k\}$ of roots of p_g differs from the sequence $\{\sigma_k\}$ of roots of p_o in the sense that $\{v_k - \sigma_k\} \in \ell^2$. This suggests that any sequence $\{v_k\}$ with this property can be synthesized with a certain feedback gain $\{\varepsilon_k\}$.

Lemma 4.1: If $\{v_k\}$ is a sequence of distinct complex numbers for which $\{v_k - \sigma_k\}$ ϵ ℓ^2 , then there exists a unique $c(\cdot)$ ϵ $L^2(0,T)$ for which the zero set of

$$p_{c}(\omega) \equiv p_{o}(\omega) + \int_{0}^{T} c(\theta)e^{\omega(T-\theta)}d\theta$$

coincides with $\{v_k\}$.

Proof: c is characterized by the moment problem

$$p_c(v_k) = 0 \implies \int_0^T c(\theta) e^{v_k(T-\theta)} d\theta = -p_0(v_k).$$

Using the properties of the exponential polynomial p_0 discussed in Section 2 together with the assumption that $\{v_k - \sigma_k\} \in \ell^2$, it is readily established that $\{p_0(v_k)\} \in \ell^2$. Moreover, it has been shown [17] that the sequence $\{e^{v_k}\}$ forms a basis for $L^2(0,T)$. Denoting the sequence biorthonormal to $\{e^{v_k}\}$ by $\{h_k(\cdot)\}$, the unique solution to the above moment problem is given by

$$c(\bullet) = -\sum_{k} p_{o}(v_{k}) \overline{h}_{k}(T-\bullet),$$

which completes the proof.

We now come to our main theorem.

Theorem 4.2: Let $\{v_k\}$ be any sequence of distinct complex numbers for which

$$\left\{\frac{v_k - \omega_k}{p'(\omega_k)b_k}\right\} \in \ell^2,$$

and let Assumption 2.4 hold true. Then there exists a unique $f \in \mathcal{H}$ for which the spectrum of $\mathcal{A}+b\otimes f$ coincides with $\{v_k\}$.

Proof. By Lemma 4.1, we may express the desired closed-loop characteristic function uniquely as

$$p_c(\omega) = p_o(\omega) + \int_0^T c(\theta)e^{\omega(T-\theta)}d\theta$$

Referring to (4.2) and (4.4), the desired feedback gain element

$$\delta = \sum_{k} f_{k} \psi_{k}$$

must be such that

$$p_g(\omega) \equiv p_o(\omega) + \int_0^T (a(\theta) - g(\theta))e^{\omega(T-\theta)}d\theta = p_c(\omega),$$

or g = a - c. Let a and c have the expansions

$$a(\cdot) = \sum_{k} a_{k} \overline{q}_{k} (T-\cdot)$$

$$c(\cdot) = \sum_{k} c_{k} \overline{q}_{k} (T-\cdot).$$

Referring to (4.2), we must have

$$f_k b_k p'(\omega_k) = a_k - c_k$$

or

$$f_k = (a_k - c_k)p'(\omega_k)^{-1}b_k^{-1}.$$
 (4.5)

This should be compared with (1.11). It remains only to show that our hypotheses ensure that $\{f_k\}$ ϵ ℓ^2 . The expansion coefficients a_k and c_k are given by

$$a_{k} = \int_{0}^{T} a(\theta)e^{\omega_{k}(T-\theta)} d\theta = p(\omega_{k}) - p_{o}(\omega_{k}) = -p_{o}(\omega_{k})$$

$$c_k = \int_0^T c(\theta) e^{\omega_k (T-\theta)} d\theta = p_c(\omega_k) - p_o(\omega_k).$$

Thus, by the intermediate value theorem,

$$a_k - c_k = -p_c(\omega_k) = -p_c(\nu_k) - p'_c(\nu_k)(\omega_k - \nu_k) = p'_c(\hat{\nu}_k)(\nu_k - \omega_k),$$

for some $\hat{\nu}_k$ on the line segment connecting ν_k and ω_k . Since $\sup_k |p_c'(\hat{\nu}_k)| < \infty$, we conclude that

$$\left\{\frac{a_k-c_k}{p'(\omega_k)b_k}\right\} \in \ell^2 \qquad \text{just in case} \quad \left\{\frac{v_k-\omega_k}{p'(\omega_k)b_k}\right\} \in \ell^2.$$

This completes the proof.

For linear hyperbolic systems of the type described in the introduction, ${\mathscr A}$ is the real operator

$$\mathscr{A} = \Lambda(x) \frac{d}{dx} + \Lambda(x),$$

with domain

$$\mathscr{D}(\mathscr{D}) = \{ \phi \in H^{1}_{2n}[0,1] : \phi^{-}(0) - D_{0}\phi^{+}(0) = 0 = \phi^{+}(1) - D_{1}\phi^{-}(1) \},$$

where H_{2n}^1 is the Sobolev space of square integrable 2n -valued functions possessing a square integrable derivative. As such the eigenfunctions of $\mathscr A$ as well as those of $\mathscr A^*$ will have the property that $\overline{\phi}_k = \phi_{-k}$. Likewise, if $\{\nu_k\}$ is a sequence of complex numbers with $\overline{\nu}_k = \nu_{-k}$, then any biorthogonal set of functions $\{h_k(\cdot)\}$ associated with the exponentials $\{e^{\lambda_k}\}$ will have the property that $\overline{h}_k = h_{-k}$. Thus, for any such sequence $\{\nu_k\}$ satisfying the hypothesis of Lemma 4.1, the function $c(\cdot)$ will be real-valued with expansion coefficients satisfying $\overline{c}_k = c_{-k}$, and hence the feedback gain element

$$f = \sum_{k} f_{k} \psi_{k} = \sum_{k} \frac{a_{k} - c_{k}}{p'(\omega_{k})b_{k}} \psi_{k}$$

constructed above will have each of its 2n component functions real valued.

We conclude this section with a brief discussion of how to use the above theory to construct a feedback gain which shifts a finite number of eigenvalues. Suppose we wish to shift $\omega_1, \omega_2, \cdots, \omega_n$ to ν_1, \cdots, ν_n . The desired closed-loop characteristic function is thus

$$p_c(\omega) = \frac{\alpha(\omega)}{\beta(\omega)} p(\omega),$$

where $\alpha(\omega) = \prod_{k=1}^{n} (\omega - v_k)$ and $\beta(\omega) = \prod_{k=1}^{n} (\omega - \omega_k)$. Thus

$$p_{c}(\omega) = p(\omega) + \frac{\alpha(\omega) - \beta(\omega)}{\beta(\omega)} p(\omega).$$

Since the degree of α - β is strictly less than n, $(\alpha\text{-}\beta)/\beta$ admits a partial fractions decomposition

$$\frac{\alpha(\omega) - \beta(\omega)}{\beta(\omega)} = \sum_{k=1}^{n} \frac{r_k}{\omega - \omega_k}.$$
 (4.6)

According to Lemma 3.3,

$$\frac{p(\omega)}{\omega - \omega_{k}} = \int_{0}^{T} \left(\Re e^{\omega_{k}} \right) (\theta) e^{\omega (T - \theta)} d\theta.$$

Thus

$$p_c(\omega) = p(\omega) + \int_0^T (\mathcal{M} \sum_{k=1}^n r_k e^{\omega_k^*}) (\theta) e^{\omega(T-\theta)} d\theta.$$

By Lemma 3.4, this is

$$p_{c}(\omega) = p(\omega) + \int_{0}^{T} \sum_{k=1}^{n} r_{k} p'(\omega_{k}) \overline{q}_{k} (T-\theta) e^{\omega(T-\theta)} d\theta$$
$$= p(\omega) + \int_{0}^{T} c(\theta) e^{\omega(T-\theta)} d\theta.$$

Thus, the desired expansion coefficients of c are given simply by $c_k = r_k p'(\omega_k)$, with the residues r_k being determined by (4.6).

5. Concluding Remarks

In this paper, we have demonstrated precisely to what extent the eigenvalues associated with a controllable pair (\mathcal{A}, b) of a certain type may be modified via continous linear state feedback. Our results parallel those of Russell [14] for the case m = 1 as explained in the introduction. Our main contributions here include a more detailed analysis of the canonical transformation and a direct method for computing feedback gains in the case of shifting a finite number of eigenvalues. All of this requires, of course, a rather detailed knowledge of the spectral structure of A. In practice, one can only obtain this information approximately by numerical computation. In this regard, our results might prove useful in providing exact solutions in special cases which could then be used to determine the accuracy of numerical computations in more general cases. It should be pointed out that it is unrealistic to assume the availability of state feeback in systems of the type discussed in this paper. Nevertheless, we feel that the results herein obtained should play a role in the development of an observer or asymptotic state estimator theory for such systems.

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1. Report No. NASA CR-172140	2. Government Accession No.	3. Recipient's Catalog No.
4. Title and Subtitle	5. Report Date June 1983	
ON THE CONTROL CANONIC	AL STRUCTURE OF A CLASS OF SCALAR	6. Performing Organization Code
7. Author(s)	8. Performing Organization Report No.	
Russell Teglas	83-18 10. Work Unit No.	
9. Performing Organization Name and		
Institute for Computer Applications in Science and Engineering Mail Stop 132C, NASA Langley Research Center		11. Contract or Grant No. NAS1-15810
Hampton, VA 23665	13. Type of Report and Period Covered Contractor Report	
Sponsoring Agency Name and Add		
National Aeronautics a Washington, D.C. 2054	14. Sponsoring Agency Code	
15. Supplementary Notes Addition Langley Technical Moni	nal support from: AFOSR-79-0018. tor: Robert H. Tolson	

16. Abstract

Final Report

In this paper, we demonstrate the equivalence of a scalar input system $\dot{x} = \mathcal{A}x + bu$, for which the eigenvalues of the generator \mathcal{A} conicide with the roots of the entire function

$$p(\omega) = e^{\omega T} + a_1 e^{\omega (T-\theta)} + \cdots + a_m + \int_0^T a(\theta) e^{\omega (T-\theta)} d\theta,$$

with the controlled scalar functional equation

$$y(t) + a_1 y(t-\theta_1) + \cdots + a_m y(t-T) + \int_0^T a(\theta) y(t-\theta) d\theta = u(t).$$

The theory of nonharmonic Fourier series is then employed to investigate the placement of eignevalues in the closed-loop system with continuous state feedback.

17. Key Words (Suggested by Author(s)) canonical form feedback nonharmonic fourier series pole assignment		18. Distribut	tion Statement	***	
		Unclassified-Unlimited Subject Category 63			
19. Security Classif, (of this report)	20. Security Classif, (of this	page)	21. No. of Pages	22. Price	
Unclassified	Unclassified	•	38	A03	

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