



# **Current Return: The Path of Least IMPEDANCE**

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## Kirchhoff's Current Law

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- All currents return to their sources following path of least **IMPEDANCE**
  - NOT always path of least resistance
- Currents do **NOT** return to ground
  - They may use ground as the return path to their source, if that is the path of least impedance

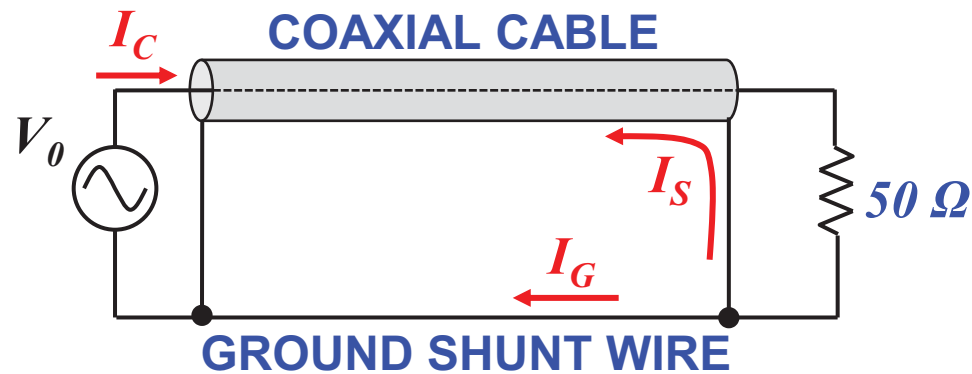
**BIG Rule of Thumb for EMC: “Follow the current”**





# Current Return Path Test Setup

- Which path will the return current take?



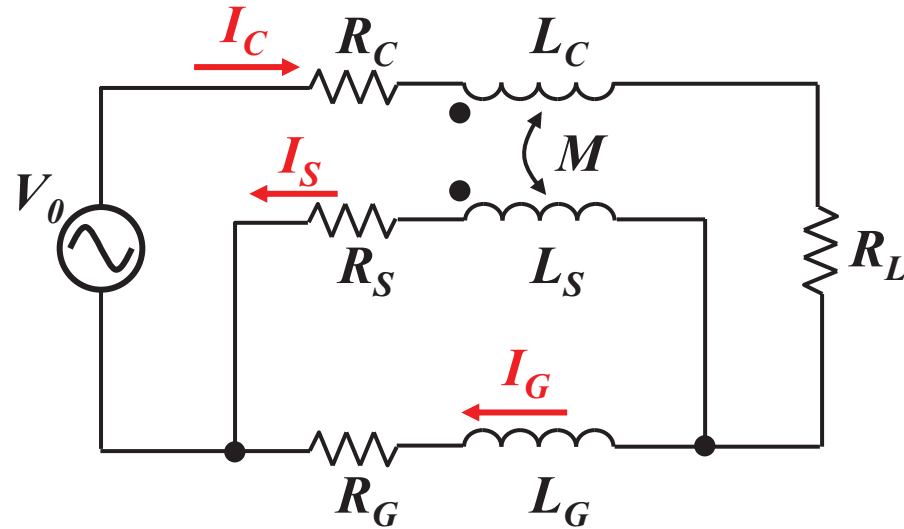
$I_C$  = center conductor current

$I_S$  = shield return current

$I_G$  = ground shunt wire return current



# Equivalent Circuit



$I_C = \text{center conductor current}$

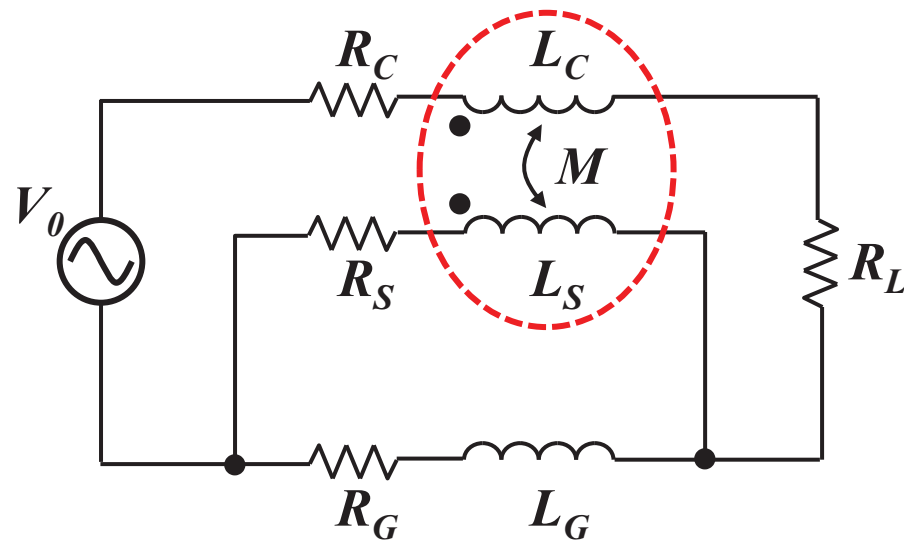
$I_S = \text{shield return current}$

$I_G = \text{ground wire return current}$

$$I_C = I_S + I_G$$



# Equivalent Circuit



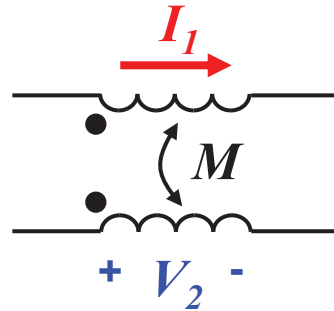
*$M = \text{mutual inductance between center conductor and shield}$*

$$*M = L_C = L_S*$$



## A Word About Mutual Inductance

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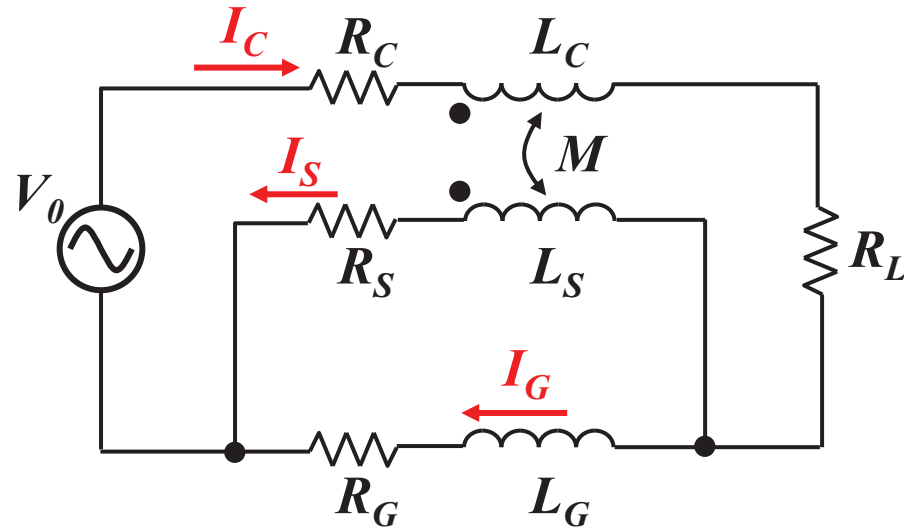


$$V_2 = M \frac{dI_1}{dt} = j\omega M I_1$$

- Mutual inductance between 2 circuits means:*
- Current in circuit 1 induces emf (potential) in circuit 2*
- Magnitude according to equation above*
  - Sign according to dot convention above*



# Circuit Analysis



$$V_0 = I_C(R_C + R_L + j\omega L_C) - I_S(j\omega M) + I_S(R_S + j\omega L_S) - I_C(j\omega M)$$

$$V_0 = I_C(R_C + R_L + j\omega L_C) - I_S(j\omega M) + I_G(R_G + j\omega L_G)$$

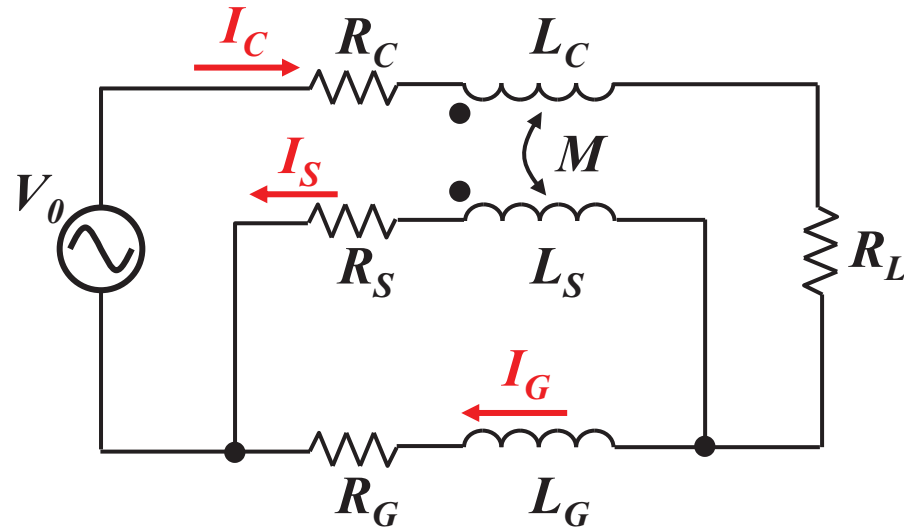
$$0 = I_S(R_S + j\omega L_S) - I_C(j\omega M) - I_G(R_G + j\omega L_G)$$

$$I_S(R_S + j\omega L_S) = I_C(j\omega M) + I_G(R_G + j\omega L_G)$$

$$I_C = I_S + I_G \quad M = L_S$$



## Circuit Analysis (cont.)



**Solving for  $I_S/I_C$ :**

$$I_S(R_S + j\omega L_S) = I_C(j\omega L_S) + (I_C - I_S)(R_G + j\omega L_G)$$

$$I_S[(R_S + R_G) + j\omega(L_S + L_G)] = I_C[(R_G + j\omega(L_S + L_G))]$$

$$\frac{I_S}{I_C} = \frac{[R_G + j\omega(L_S + L_G)]}{[(R_S + R_G) + j\omega(L_S + L_G)]}$$

**Solving for  $I_G/I_C$ :**

$$(I_C - I_G)(R_S + j\omega L_S) = I_C(j\omega L_S) + I_G(R_G + j\omega L_G)$$

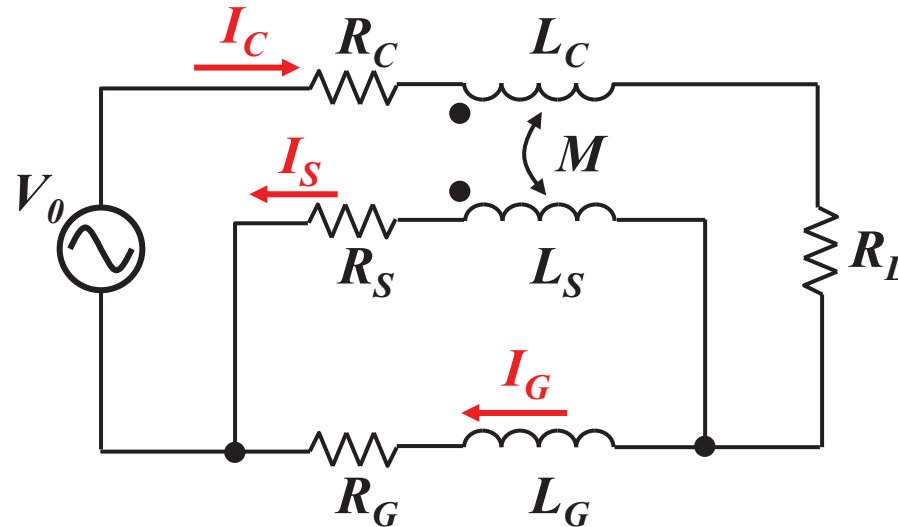
$$I_C R_S = I_G[(R_S + R_G) + j\omega(L_S + L_G)]$$

$$\frac{I_G}{I_C} = \frac{R_S}{[(R_S + R_G) + j\omega(L_S + L_G)]}$$





# Circuit Analysis (cont.)



**Shield return current:**

$$\frac{I_S}{I_C} = \frac{[R_G + j\omega(L_S + L_G)]}{[(R_S + R_G) + j\omega(L_S + L_G)]}$$

**At low frequencies:**

$$\frac{I_S}{I_C} \approx \frac{R_G}{R_S + R_G}$$

*Resistive divider*

**Ground wire return current:**

$$\frac{I_G}{I_C} = \frac{R_S}{[(R_S + R_G) + j\omega(L_S + L_G)]}$$

**At low frequencies:**

$$\frac{I_G}{I_C} \approx \frac{R_S}{R_S + R_G}$$

**At high frequencies:**

$$\frac{I_S}{I_C} \approx \frac{j\omega(L_S + L_G)}{j\omega(L_S + L_G)} \approx 1$$

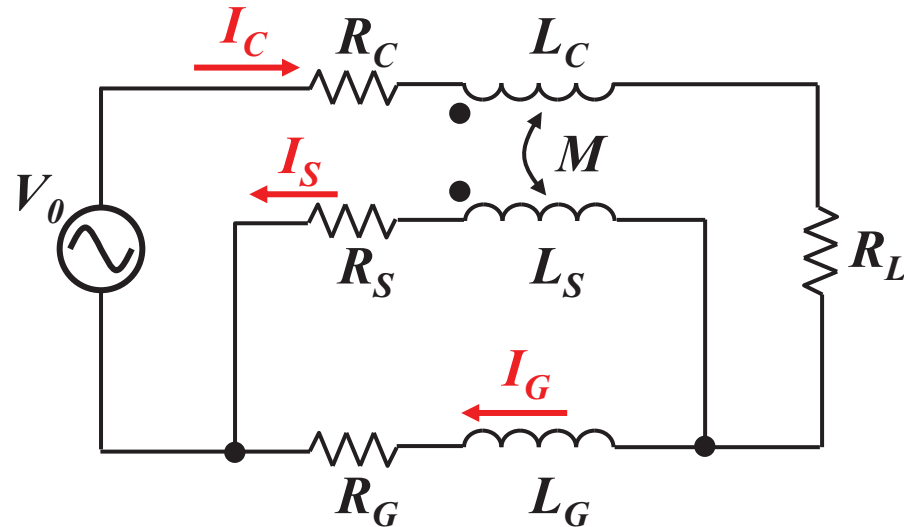
*Virtually all current returns on shield*

**At high frequencies:**

$$\frac{I_G}{I_C} \approx \frac{R_S}{j\omega(L_S + L_G)} \approx 0$$



## Circuit Analysis (cont.)



**Ratio of shield current to ground wire current:**

$$\frac{I_S}{I_G} = \frac{[R_G + j\omega(L_S + L_G)]}{R_S}$$

**$I_S = I_G$  when:**

$$\sqrt{R_G^2 + [2\pi f_c(L_S + L_G)]^2} = R_S$$

$$f_c = \frac{\sqrt{R_S^2 - R_G^2}}{2\pi(L_S + L_G)}$$



## Test #1 Setup Parameters

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### *Shield return path:*

$$R_S \text{ (measured)} = 96 \text{ m}\Omega$$

$$L_S / l = 0.25 \text{ }\mu\text{H/m (from RG-58 coaxial cable datasheet)}$$

$$l = 4.8 \text{ m}$$

$$L_S = 1.2 \text{ }\mu\text{H}$$

### *Ground wire return path:*

$$R_G \text{ (measured)} = 15 \text{ m}\Omega$$

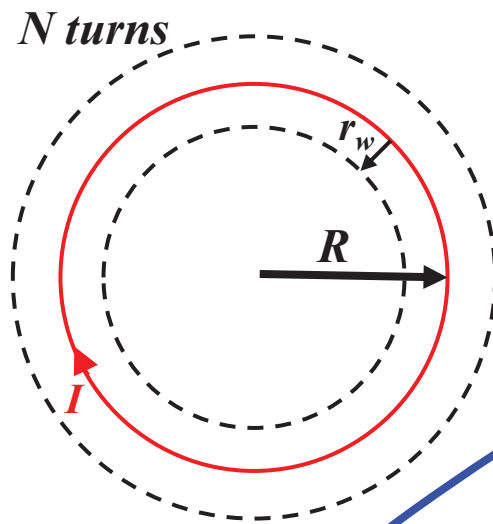
$$L_G = 24.4 \text{ }\mu\text{H (next slides)}$$





# Inductance of Circular Loop

- From Missouri University of Science and Technology inductance calculator
  - <http://emclab.mst.edu/inductance/>



$$L_{loop} \approx N^2 \mu_r \mu_0 R \left( \ln \frac{8R}{r_w} - 2 \right)$$

$$N = \sim 5.5$$

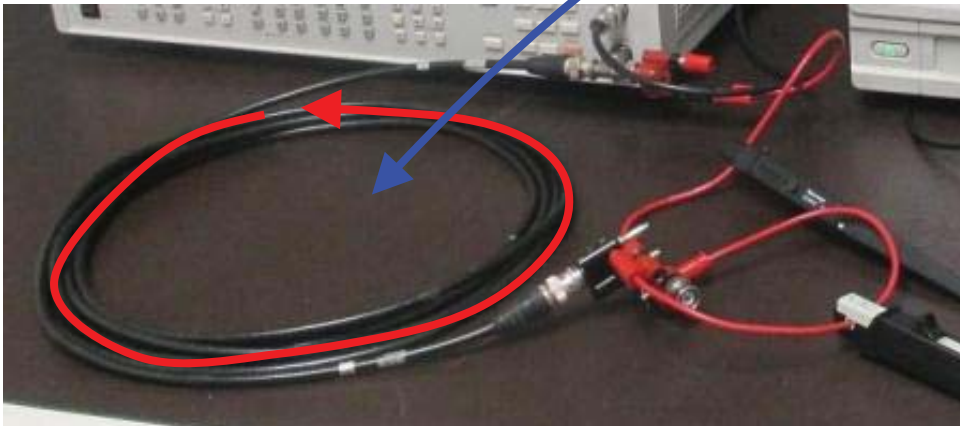
$$\mu_r = 1$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

$$R = 6 \text{ inches} = 15 \text{ cm}$$

$$r_w = 0.25 \text{ cm (shield radius)}$$

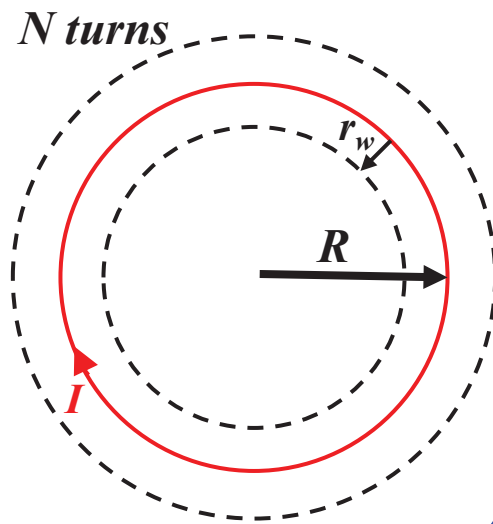
$$L_{coil} = \sim 23.8 \mu\text{H}$$





## Inductance of Circular Loop (cont.)

- Coiled cable in series with loop completing connections to signal generator:



$$L_{loop} \approx N^2 \mu_r \mu_0 R \left( \ln \frac{8R}{r_w} - 2 \right)$$

$$N = 1$$

$$\mu_r = 1$$

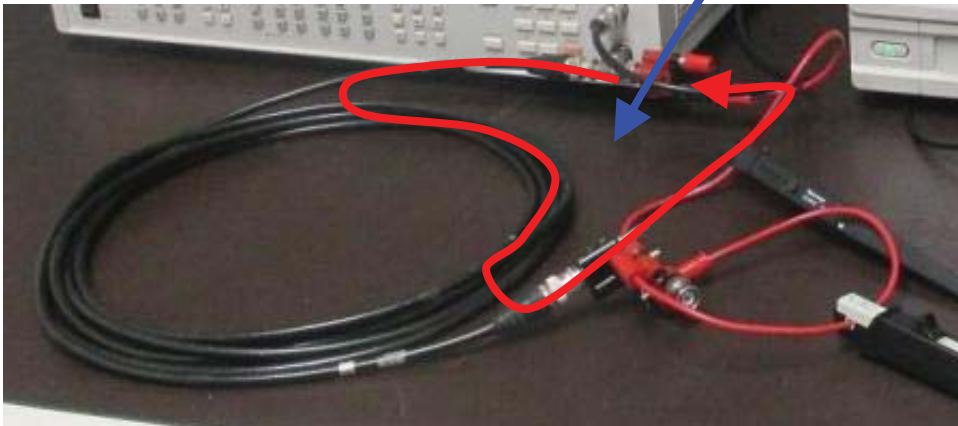
$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

$$R_{eff} = 12 \text{ cm}$$

$$r_w = 0.25 \text{ cm (shield radius)}$$

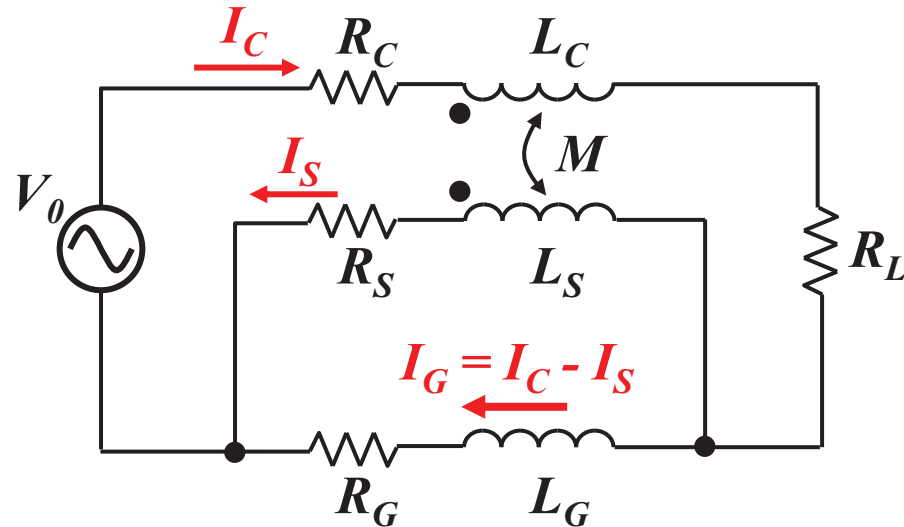
$$L_{gen} = \sim 0.6 \mu\text{H}$$

$$L_G = L_{coil} + L_{gen} = 24.4 \mu\text{H}$$

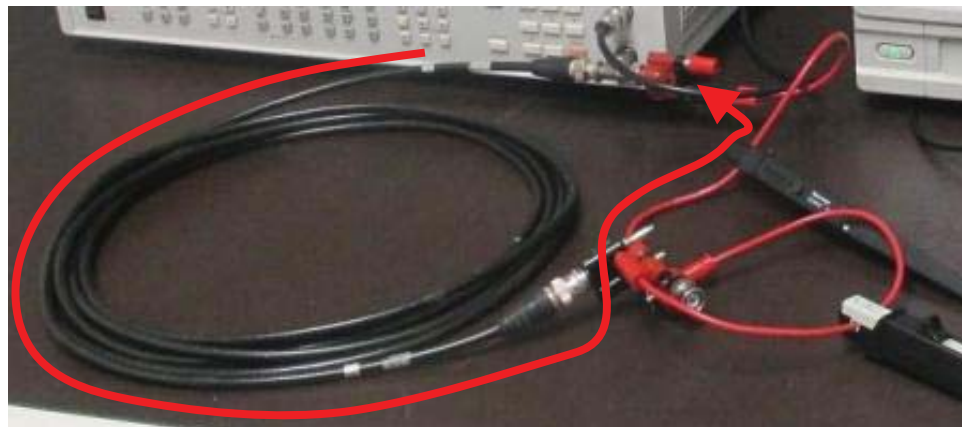




## Why Do We Use the Shield Radius?

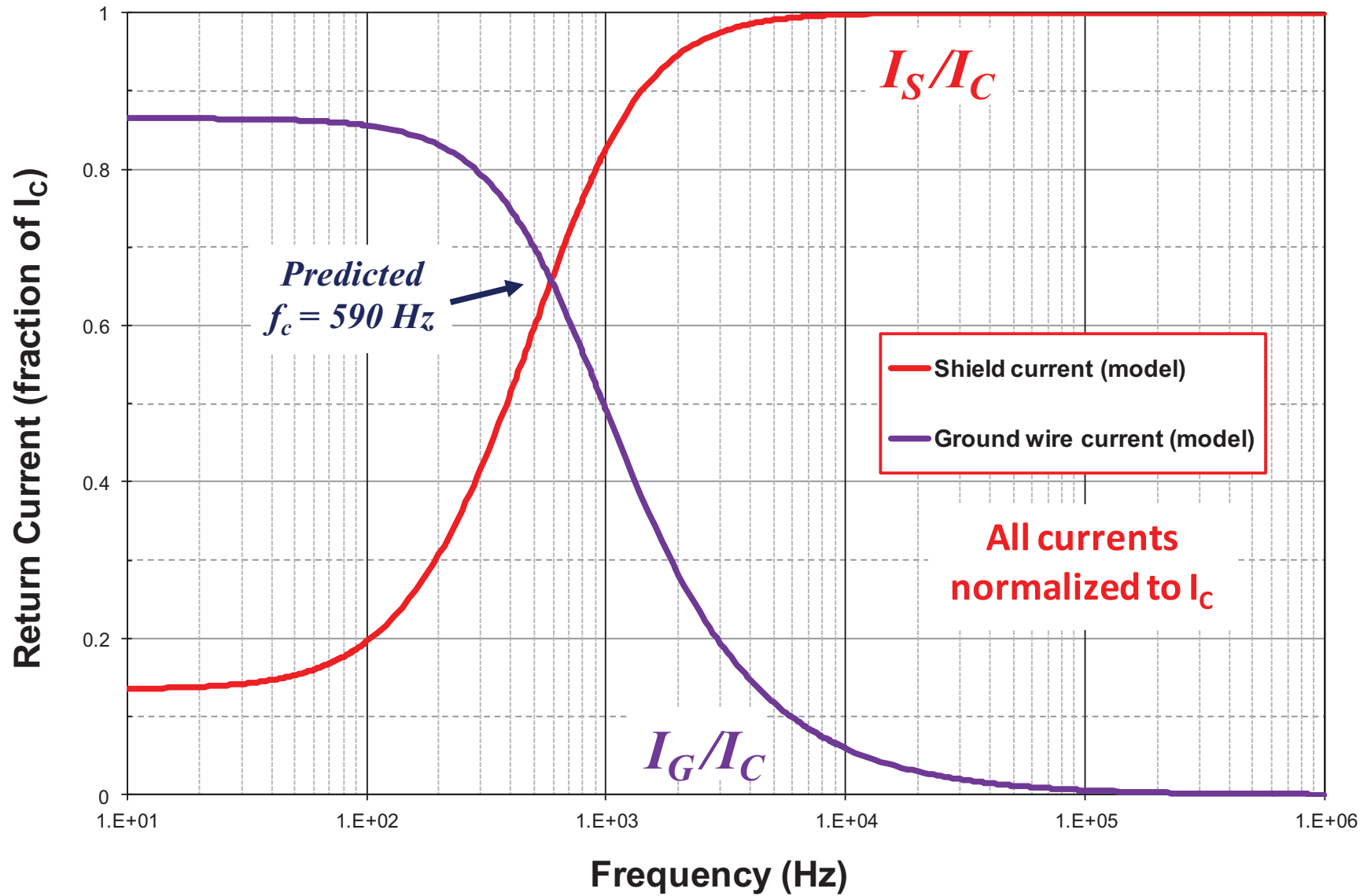


*We use the shield radius because the current in the large loop is the net current from the coax, i.e. that gets past the shield*



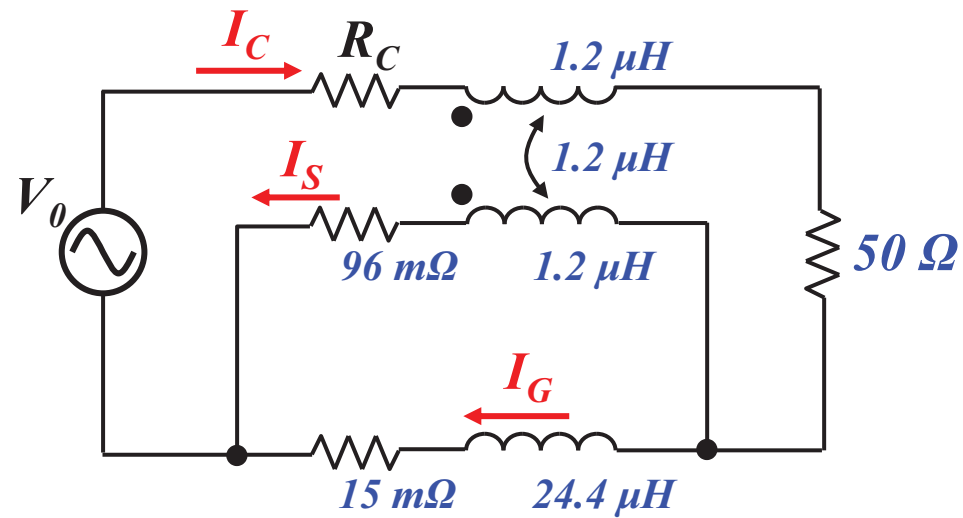


# Return Currents – Model for Test #1





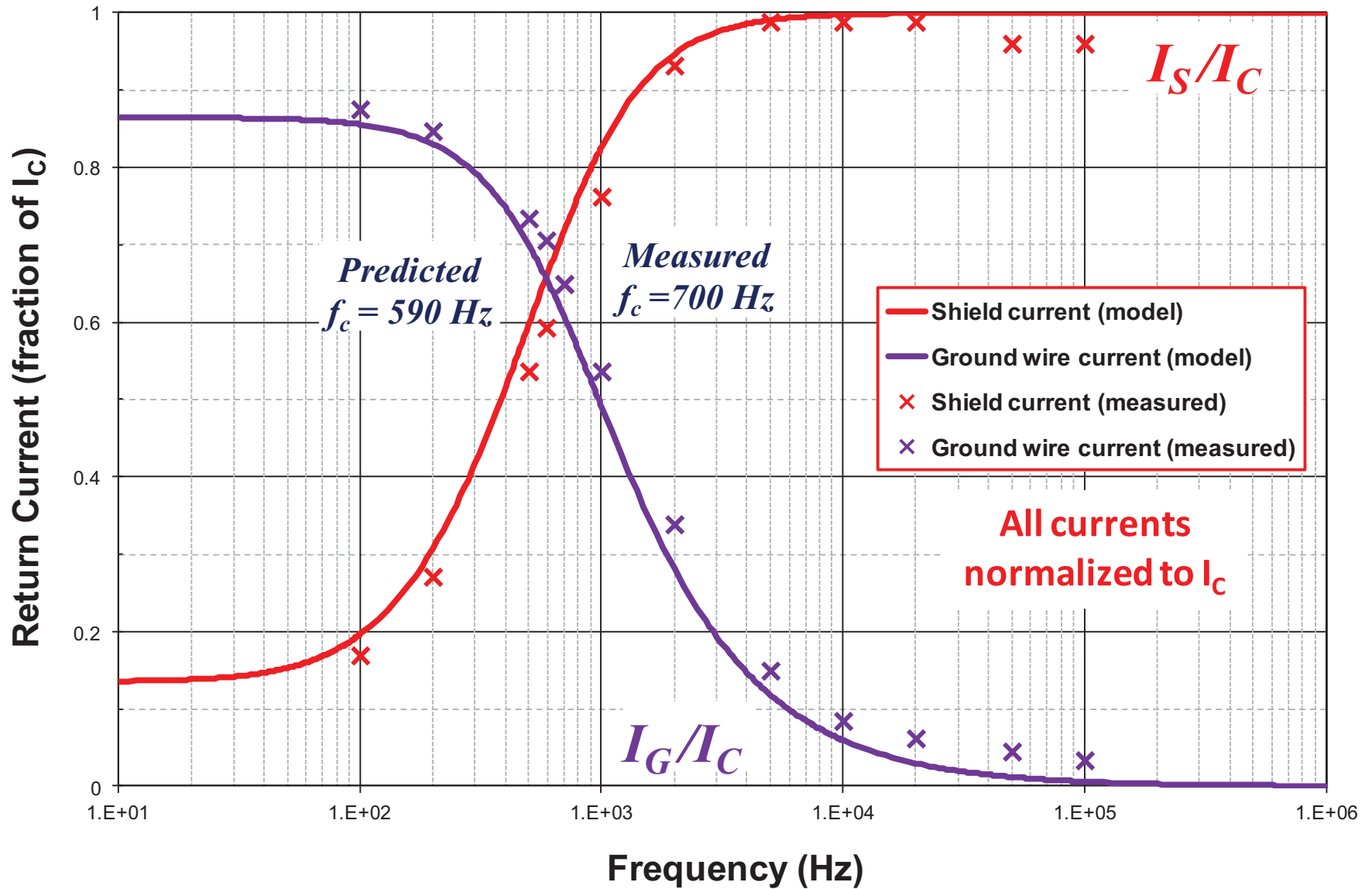
# Updated Equivalent Circuit





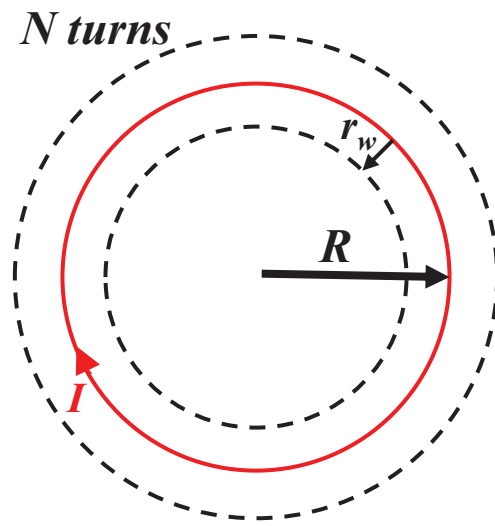


# Test #1: Measured vs. Model





## Test #2: 3.5-Turn Loop



$$L_{loop} \approx N^2 \mu_r \mu_0 R \left( \ln \frac{8R}{r_w} - 2 \right)$$

$$N = \sim 3.5$$

$$\mu_r = 1$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

$$R = 9 \text{ inches} = 23 \text{ cm}$$

$$r_w = 0.25 \text{ cm (shield radius)}$$

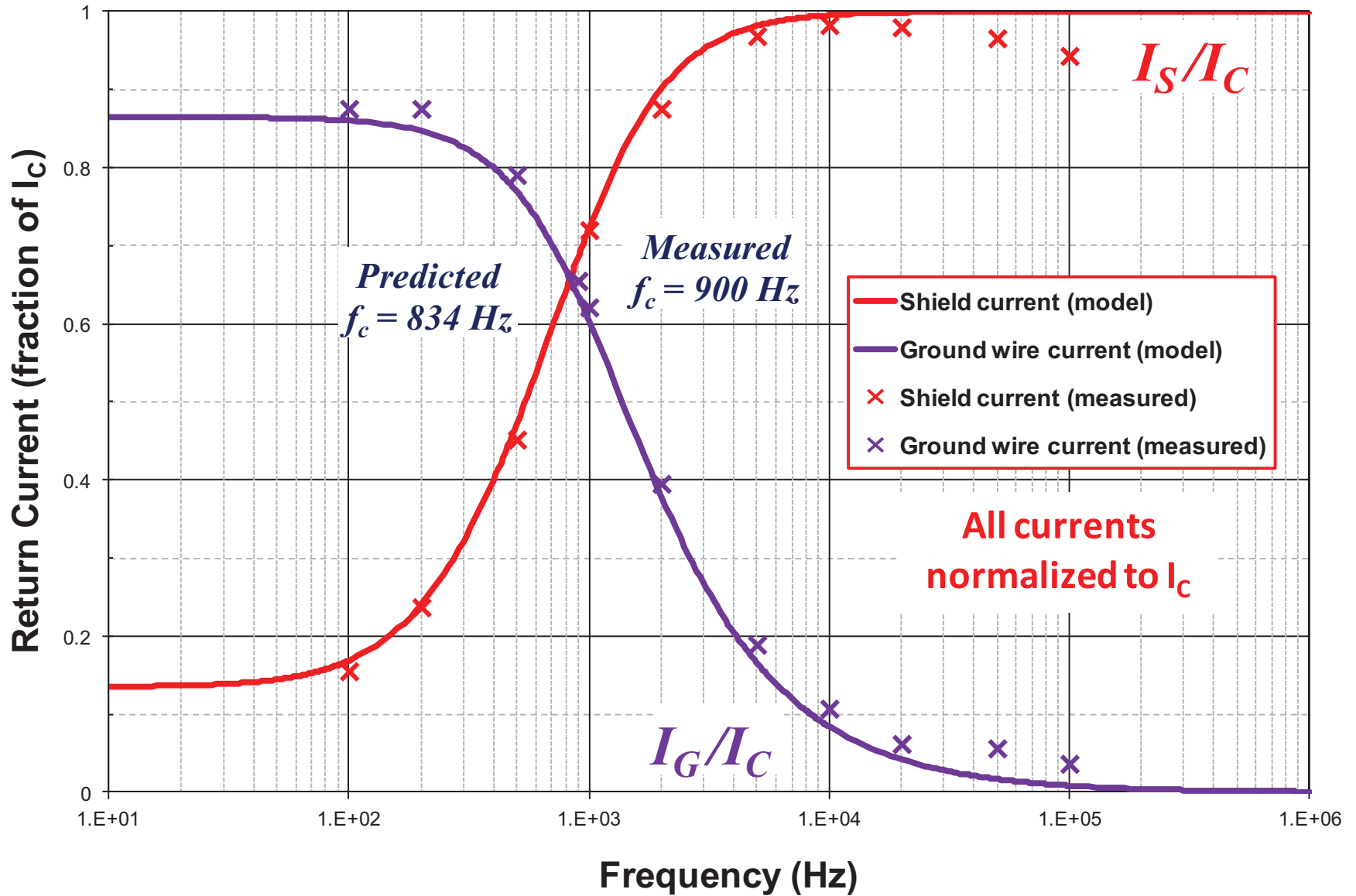
$$L_{coil} = \sim 16.3 \mu\text{H}$$

$$L_{gen} = \sim 0.6 \mu\text{H (same as before)}$$

$$L_G = L_{coil} + L_{gen} = 16.9 \mu\text{H}$$

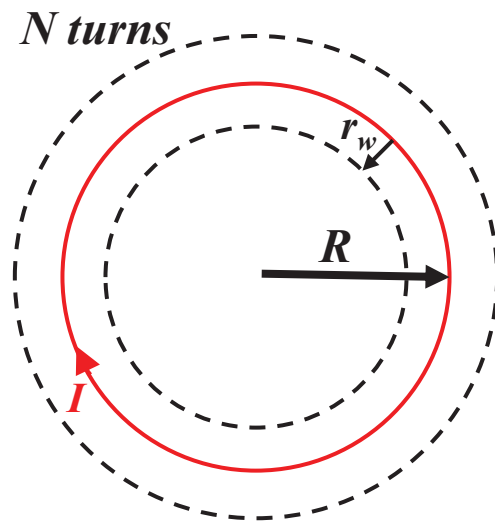


# Test #2: Measured vs. Model





## Test #3: 1-Turn Loop



$$L_{loop} \approx N^2 \mu_r \mu_0 R \left( \ln \frac{8R}{r_w} - 2 \right)$$

$$N = 1$$

$$\mu_r = 1$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

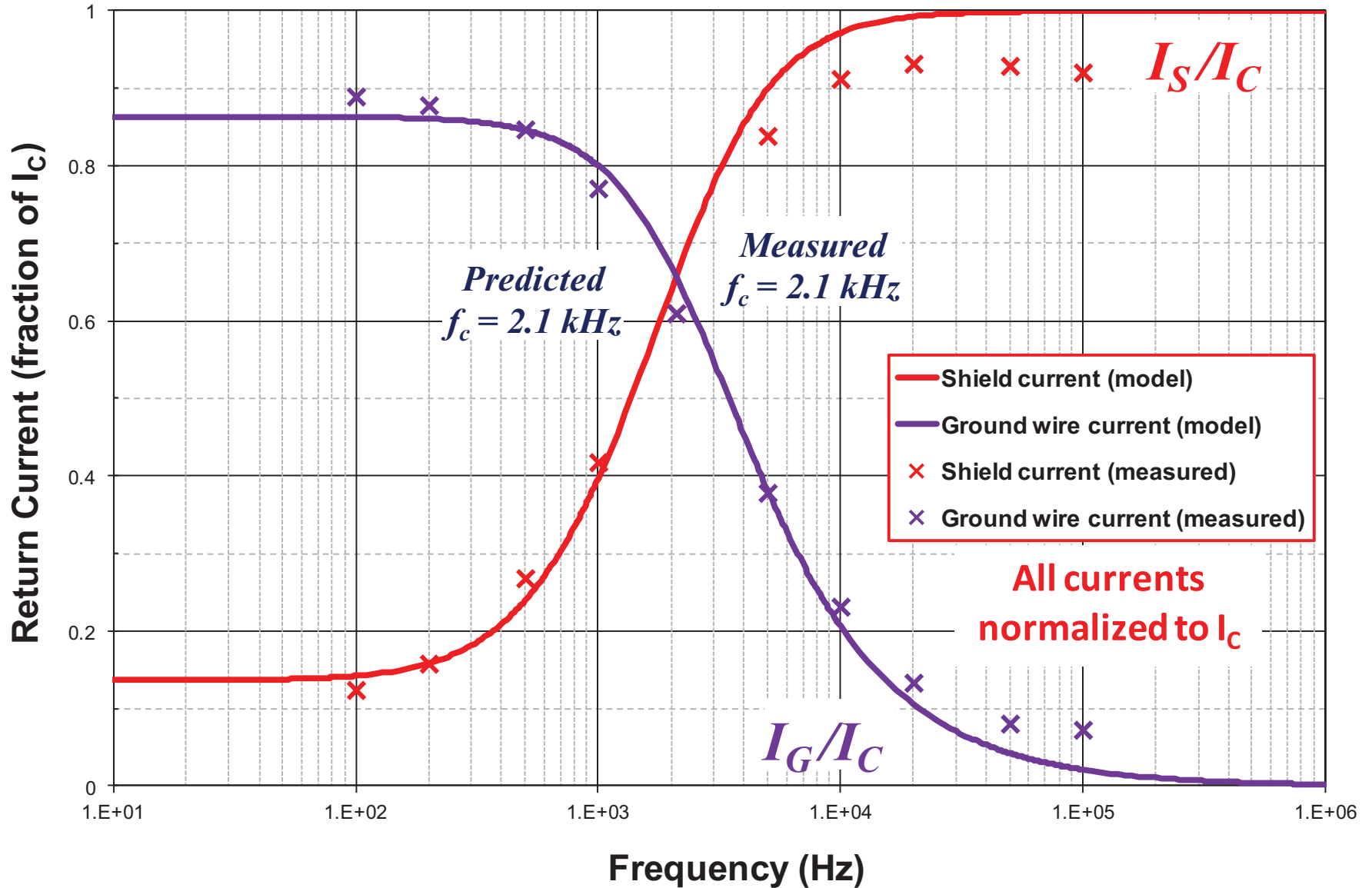
$$R_{eff} = 5 \text{ m} / 2\pi = \sim 80 \text{ cm}$$

$$r_w = 0.25 \text{ cm (shield radius)}$$

$$L_G = 5.9 \mu\text{H}$$

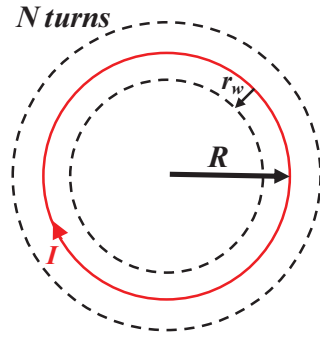


# Test #3: Measured vs. Model





## Why Doesn't the Ground Wire Current Go To Zero?



$$L_{loop} \approx N^2 \mu_r \mu_0 R \left( \ln \frac{8R}{r_w} - 2 \right)$$



$$N = 1$$

$$\mu_r = 1$$

$$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$$

$$R = \sim 5 \text{ cm}$$

$$r_w = 0.05 \text{ cm}$$

$$L_G = \sim 300 \text{ nH}$$

**THE INDUCTANCE OF THIS LOOP LIMITS THE SHIELD CURRENT AT HIGHER FREQUENCIES**



## Observations/Summary

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- **Measured data shows good agreement with model**
  - At low frequencies, current forms resistive divider with all available paths
  - At high frequencies, impedance is dominated by inductance (loop area)
- **High inductance (large loop areas) increases likelihood of ground bounce and magnetic coupling (crosstalk) between circuits**
- **Practical implications for circuit and system design:**
  - Provide deliberate low inductance (small loop area) return paths for high frequency currents
  - Cabling: twisted pairs, coax
  - PC board: return (-) trace immediately adjacent to send (+) trace
  - Do NOT route traces over splits in ground plane
  - Video type signals
    - Video signals tend to be “bursty”; mix of high and low frequency content
    - High frequency content will return on the shield; low frequency content will not
    - Shield isn't a shield at low frequencies, and thus the low frequency video signal mixes in with ground noise which pollutes the video signal



## Questions/Comments

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- **Contact:**

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