



A Novel Reconfigurable GaN Based Fully Solid-State Microwave Power Module for Communications/Radar Applications

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Abstract

The paper presents as a proof-of-concept (POC) the design, integration, and performance of a novel reconfigurable S-/X-band, GaN based, fully solid-state microwave power module (SSMPM) to enable miniaturization of the overall RF system. The SSMPM includes diplexers, pre-amplifiers, a multistage medium power amplifier, SPDT switches, and CW/Pulsed high-power amplifiers. These components are synergistically integrated such that a single SSMPM is capable of being dynamically reconfigured to function as a CW S-/X-band amplifier for TT&C/science data downlink and as a pulsed X-band amplifier for remote sensing radar onboard a planetary exploration spacecraft. The POC SSMPM is capable of delivering P_{sat} of 39 dBm (8 W CW) at S-band, P_{sat} of 46 dBm (40 W CW) at X-band, and P_{sat} of >50 dBm (>100 W Pulsed) at X-band.

1.0 Introduction

Historically, the term microwave power module (MPM) is associated with a small fully integrated self-contained radio frequency (RF) amplifier that combines both solid-state and microwave vacuum electronics technologies (Figure 1) (Ref. 1). In this paper, we present the research and development of a novel fully solid-state microwave power module (SSMPM), which is distinctly different from the above MPMs (Ref. 2). The SSMPM advances the state-of-the-art in spacecraft transmitters. Our effort leverages from the recent advances in RF wide bandgap semiconductor (WBGs) crystal growth, wafer preparation, device/circuit fabrication technologies and reliability studies that have resulted from the investment that Defense Advanced Research Projects Agency (DARPA) has made with the U.S. industry (Ref. 3). As a result of DARPA investments, the past few years have witnessed significant progress in the development of gallium nitride (GaN) high electron mobility transistor (HEMT) on silicon carbide (SiC) substrate based monolithic microwave integrated circuits (MMICs) for RF power applications. The main advantages of GaN HEMT on SiC substrate is that it has power density as high as 12.2 W/mm at X-band (Ref. 4). Thus, for a desired output power a GaN MMIC is smaller in size compared to a gallium arsenide (GaAs) MMIC. In addition, SiC substrate has 8 to 10

times higher thermal conductivity than a GaAs substrate (Ref. 5), which allows operating at higher junction temperature and also enhances thermal reliability by more efficiently conducting waste heat away from the junction.

A typical payload on a near-Earth and deep space planetary exploration spacecraft includes S-band system for telemetry, tracking, and command (TT&C) and a X-band or higher frequency (Ka-band) telecommunication system for down linking data from science instruments. The down-link frequency ranges allocated by the International Telecommunication Union (ITU) for use in deep space and near-Earth research are S-band: 2.29 to 2.3 GHz and 2.2 to 2.29 GHz, respectively and X-band: 8.4 to 8.45 GHz and 8.45 to 8.5 GHz, respectively. The role of the TT&C system is receiving commands and downlinking spacecraft house-keeping data. Typical science instruments are scatterometers, radiometers, and radar for synthetic aperture imagers. The role of the telecommunication system is to down-link science data acquired by these instruments. The current state of practice uses two separate S-band and X-band amplifiers in each of the above systems. However, due to the push for developing small satellites with enhanced system capabilities/performance at lower cost, it is advantageous to develop a single wideband, reconfigurable high-power, high-efficiency SSMPM that can operate at multiple frequency bands depending on the need at any given time. Additionally, a single dynamically reconfigurable amplifier enables miniaturization of the overall RF system. Innovations in compound semiconductor materials, devices, and circuits to increase the functionality and reconfigurability of RF systems are reported in References 6 and 7.

In this paper, we present as a proof-of-concept (POC) the design, integration, and performance of a novel reconfigurable S-/X-band GaN based fully solid-state MPM with a view to miniaturize the overall RF system. The module synergistically integrates diplexers, pre-amplifiers, multistage medium power amplifiers (MPAs), single pole, double throw (SPDT) switches, and CW/Pulsed high power amplifiers (HPAs) with a voltage sequencer, a DC blanking controller, and a low voltage electronic power conditioner. The POC SSMPM operates at both S-band and X-band to serve multiple roles. The SSMPM can be reconfigured to deliver P_{sat} of 39 dBm (8 W CW) at S-band, P_{sat} of 46 dBm (40 W CW) at X-band, and P_{sat} of >50 dBm (>100 W Pulsed) at X-band.

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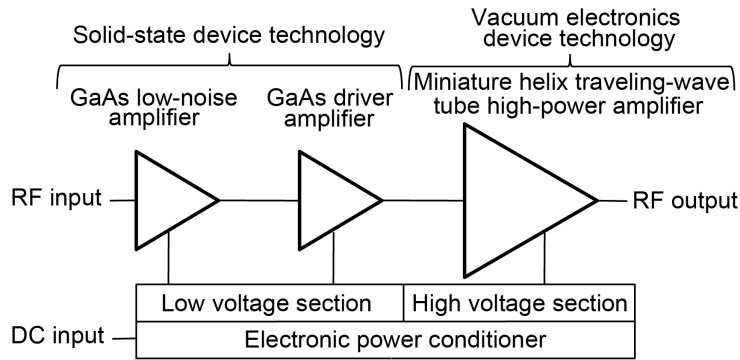


Figure 1.—Schematic of a conventional microwave power module (MPM) based on both solid-state and microwave vacuum electronics technologies.

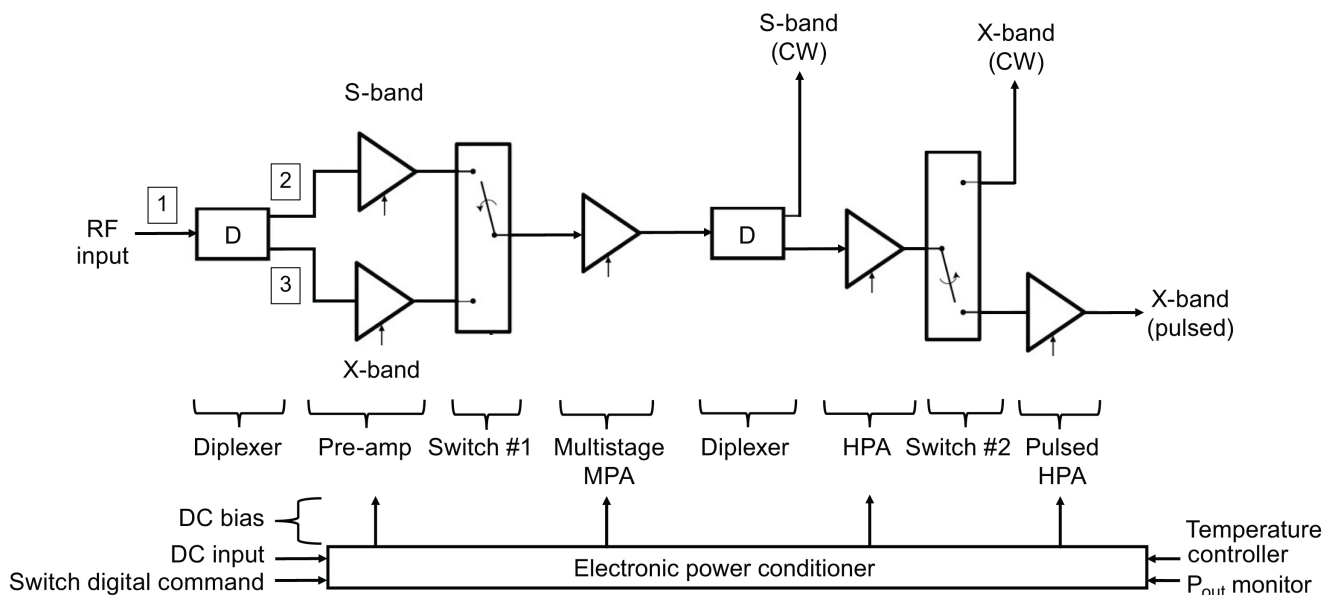


Figure 2.—Schematic of a reconfigurable fully solid-state microwave power module (SSMPM) based on high power GaN MMIC Power Amplifiers.

2.0 Solid-State Microwave Power Module

2.1 Module Design and Mode of Operation

The SSMPM layout is schematically illustrated in Figure 2. The typical frequencies at S-band and X-band for TT&C and telecommunications are 2.2 and 8.4 GHz, respectively. At the input, a diplexer (D) selects either the S-band or the X-band signal to be processed. The two pre-amplifier stages are GaAs based MMIC amplifiers that amplify the S-band and the X-band output signals from the diplexer respectively, to a level appropriate to drive the common multistage MPA. The switch is a nonreflective silicon single-pole, double throw (SPDT) device. The multistage MPA is high efficiency GaN HEMT PAs with output power (P_{out}) sufficient to drive the HPA stage.

The HPAs are also high-efficiency GaN HEMT PAs that provide CW or pulsed power. As opposed to conventional amplifiers with external input/output matching networks, which take up relatively large area of dielectric substrate for microwave circuitry, the MPA and the HPA are designed with internally matched transistors that enables miniaturizing the overall SSMPM (Ref. 8).

An electronic power conditioner (EPC) provides the gate and drain voltages and currents for the above amplifier stages. The EPC is a DC-to-DC power converter that transforms the spacecraft bus voltage typically in the range of +21 V and +35 V into regulated voltages required by the amplifier stages. In addition, a DC power management circuit is included to manage the correct power-up and power-down sequence. That is to ensure that the negative gate voltages are applied before the positive drain voltages are applied to turn the amplifiers ON.

Furthermore, a DC blanking control is also provided to quickly turn the amplifiers OFF if a fault condition arises. Moreover, a RF output monitor such as a temperature sensor or a detector/reference diode pair is located near or on the high-power GaN die, in the output stage, to monitor for an over temperature condition. The detector/reference diode pair also monitors the RF output power level. The packaged SSMPM unit will be conduction cooled.

3.0 Measured Results

In this section, the measured performance of the diplexer, pre-amplifiers, SPDT switch, MPA, and HPA under CW and pulsed operating conditions are presented. The MPA and the HPA were maintained at 25 °C base plate temperature during characterization by attaching them to Peltier-thermoelectric cold plate coolers.

3.1 Diplexer

The diplexer is Marki Microwave Model DPX-4 and is housed in a miniature connectorized package. The measured insertion loss from the input to the output low-pass (LP) (S-band) and the high-pass (HP) (X-band) coaxial ports are on the order of 0.8 and 1.0 dB at the S-band and X-band, respectively. The passband return loss and the common port return loss are better than 10 dB. The isolation is better than 25 dB. These results are presented in Figure 3.

3.2 Pre-Amplifiers

The two Avantek pre-amplifiers, model AWT-6035 operating from 2 to 6 GHz and model AMT-12436 operating from 7 to 12.4 GHz, each provides a gain of about 50 dB with output power and gain presented as a function of frequency in Figure 4.

3.3 SPDT Switch

The switch is an Analog Devices Model HMC1118 and is a non-reflective SPDT type that operates over a wideband (9 kHz to 13 GHz), which allows reconfiguring the SSMPM to function at either S- or X-band. In addition, the switch is internally matched to 50 ohms at the RF input port and the RF output ports and hence requires no external matching components. The HMC1118 is housed inside a 3 by 3 mm surface mount package. The measured insertion loss of the

switch when the two inputs are sequentially turned ON is on the order of 0.6 and 1.4 dB at S-band and X-band, respectively (Figure 5(a)). The hatched areas in Figure 5 indicate the down-link frequency ranges for use in deep space and near-Earth research. The measured isolation between the two output ports of the switch is greater than 50 and 40 dB at S-band and X-band, respectively (Figure 5(b)).

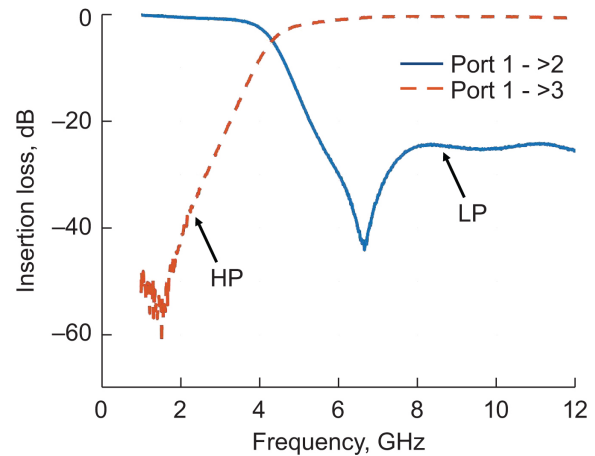


Figure 3.—Measured diplexer insertion loss at the low-pass (S-band) and high-pass (X-band) coaxial ports.

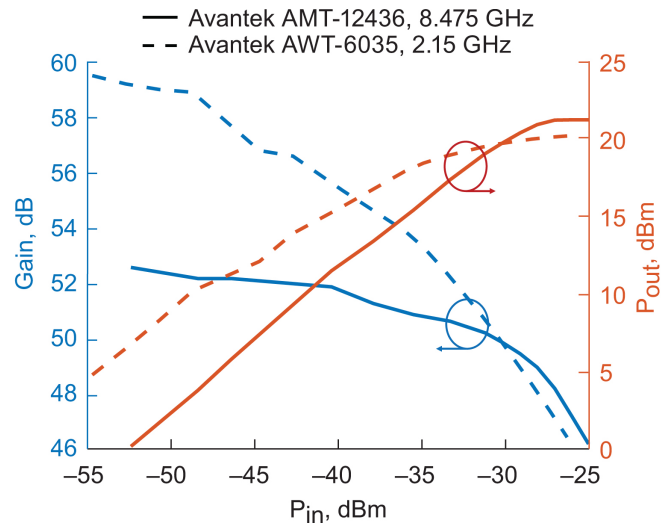
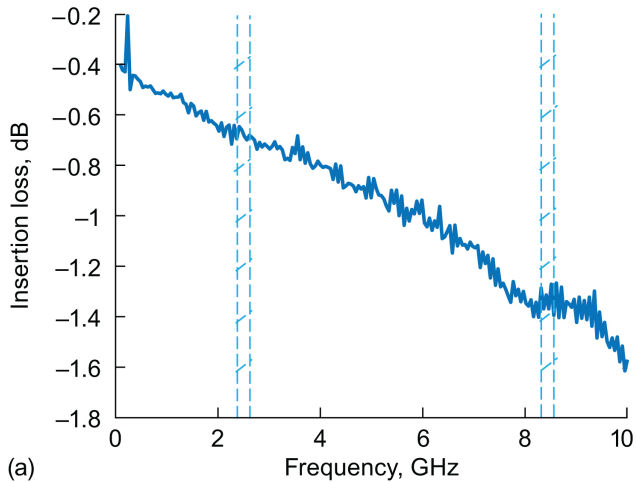
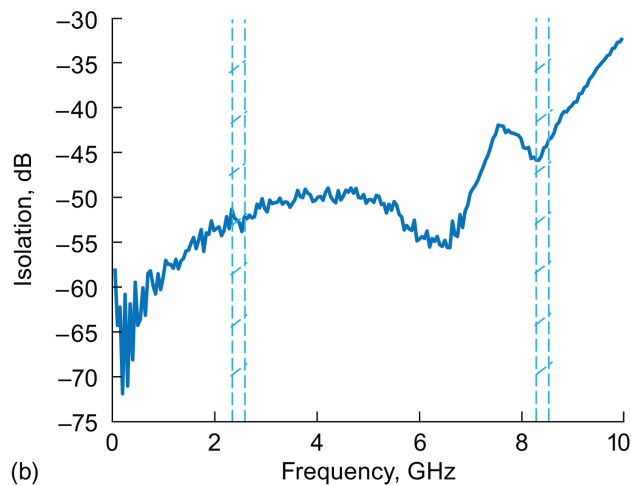


Figure 4.—Measured output power and gain as a function of the input drive at S-band (2.15 GHz) for Avantek AWT-6035 with $V_D = 12\text{ V}$, $I_{DQ} = 310\text{ mA}$ and at X-band (8.475 GHz) for Avantek AMT-12436 with $V_D = 12\text{ V}$, $I_{DQ} = 350\text{ mA}$.



(a)

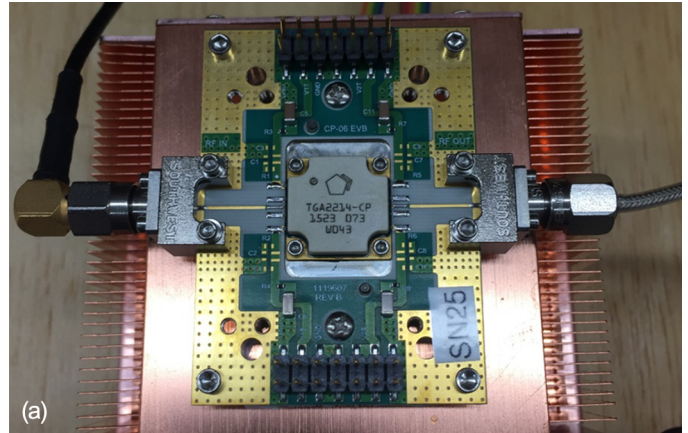


(b)

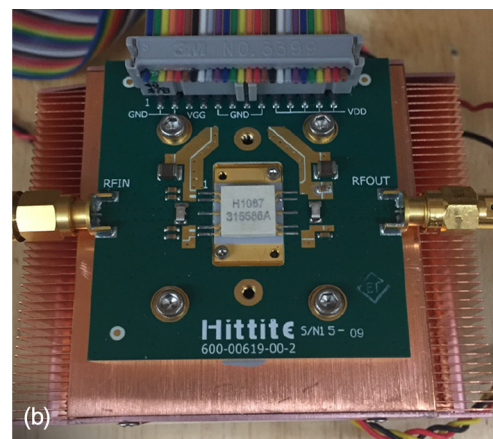
Figure 5.—(a) Measured insertion loss as a function of the S-band and X-band frequencies, and (b) Measured isolation as a function of the frequency, for the Analog Devices HMC1118 SPDT switch with $V_{DD} = 3.3$ V, $V_{SS} = -2.5$ V, and $V_{CTRL} = 0$ V/3.3 V.

3.4 Wideband GaN Multistage Medium Power Amplifier (MPA)

The 1st stage of the wideband MPA is built with a Qorvo Model TGA2214-CP MMIC, which has $0.15 \mu\text{m}$ gate transistors fabricated on GaN-on-SiC substrate and operates from 2 to 18 GHz. The input and output RF ports of the MMIC have integrated DC blocking capacitors and are fully matched to 50 ohms. Furthermore, the MMIC package has pure copper base offering superior thermal management (Figure 6(a)). The 2nd stage of the MPA is built with Analog Devices Model



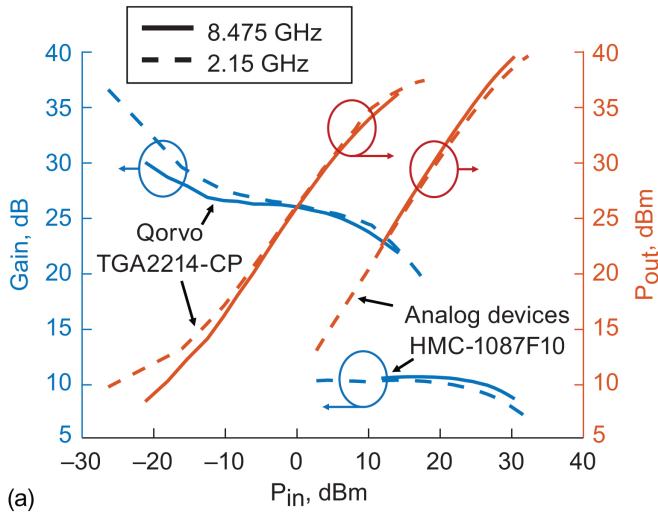
(a)



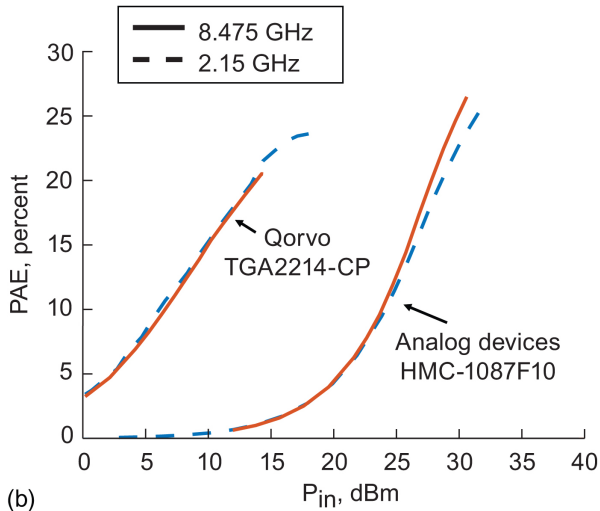
(b)

Figure 6.—(a) 1st stage, and (b) 2nd stage of the GaN wideband multistage MPA mounted on copper heat sinks with cooling fans during characterization.

HMC1087F10 GaN MMIC that operates from 2 to 20 GHz. The MMIC input and output ports are internally matched to 50 ohms. Moreover, the MMIC package body material is copper tungsten ($15\text{Cu}85\text{W}$) (Figure 6(b)). The measured P_{out} and gain of each stage at S-band (2.15 GHz) and X-band (8.475 GHz) as a function of input drive power are presented in Figure 7(a). The corresponding measured PAEs are presented in Figure 7(b). The P_{sat} of the 1st stage is 37 dBm, the corresponding gain is 20 dB, and the PAE is 25 percent. The P_{sat} of the 2nd stage is 39.5 dBm, the corresponding gain is 8.8 dB, and the PAE is 27 percent. The measured P_{out} and gain of the 1st and the 2nd stage as a function of the S-band and X-band frequencies are presented in Figure 8(a) and Figure 9(a), respectively. The corresponding measured PAEs are presented in Figure 8(b) and Figure 9(b). The P_{out} , gain, and PAE are nearly uniform across the frequency ranges of interest.

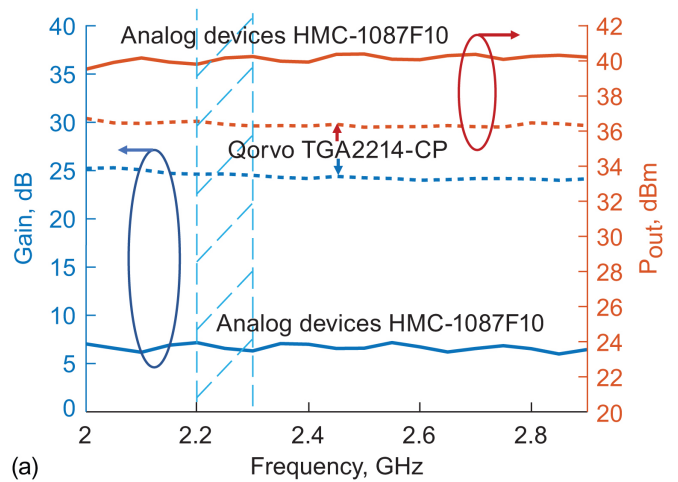


(a)

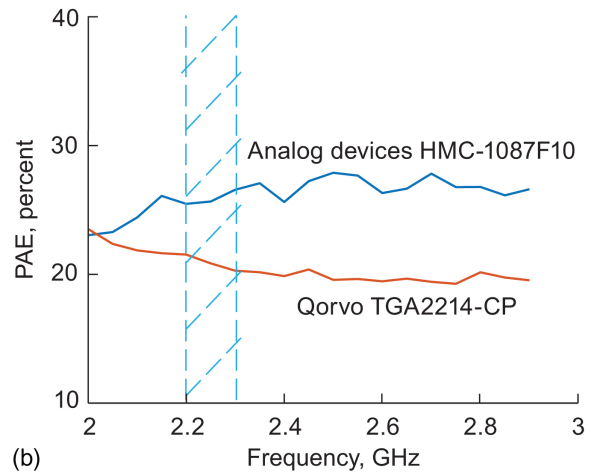


(b)

Figure 7.—(a) Measured output power and gain as a function of the input drive, and (b) Measured PAE as a function of the input power, for the MPA 1st stage with TGA2214-CP: $V_D = 22$ V, $V_G = -2.3$ V, $I_{DQ} = 600$ mA and 2nd stage with HMC1087F10: $V_D = 28$ V, $V_G = -2.3$ V, and $I_{DQ} = 850$ mA.

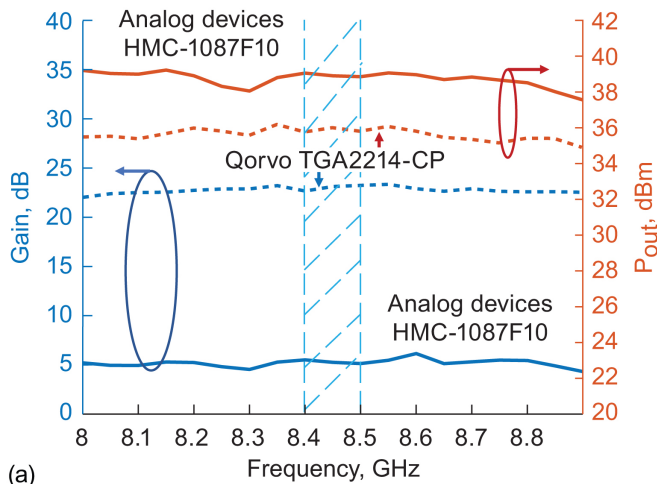


(a)

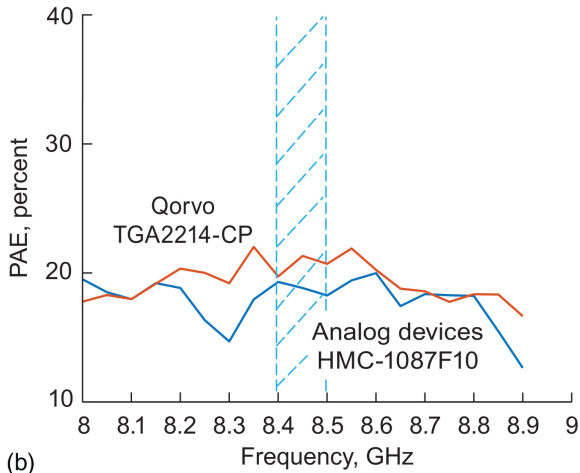


(b)

Figure 8.—(a) Measured output power and gain, and (b) Measured PAE as a function of the S-band frequencies, for the MPA 1st and 2nd stage with input drive power in the range of 11.38 to 12.62 dBm and 32.23 to 34.03 dBm, respectively.



(a)

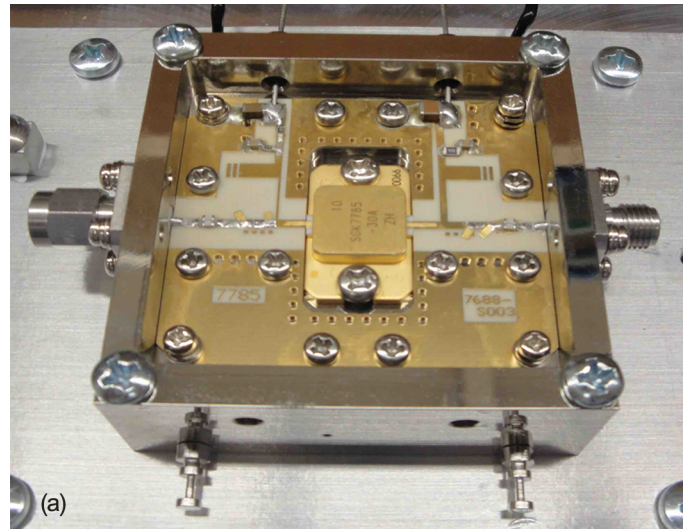


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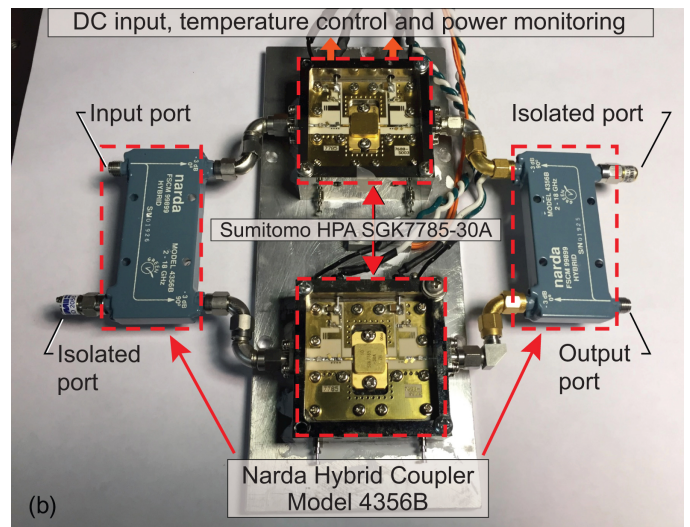
Figure 9.—(a) Measured output power and gain, and (b) Measured PAE as a function of the X-band frequencies, for the MPA 1st and 2nd stage with input drive power in the range of 12.4 to 13.49 dBm and 33.05 to 34.06 dBm, respectively.

3.5 GaN CW High-Power Amplifier (HPA)

The CW HPA stage is configured either as a single Sumitomo Model SGK7785-30A GaN HEMT based MMIC amplifier (Figure 10(a)) or as two SGK7785-30A amplifiers in a balanced configuration. The MMIC amplifier is internally matched to 50 ohms. The balanced amplifier, with topology shown in Figure 10(b) employs hybrid couplers (Narda Model 4356B, 2 to 18 GHz) at the input and output to divide and combine the power, respectively. The measured P_{out} and gain and PAE as a function of the input drive power at X-band (8.475 GHz) are presented in Figure 11(a) and (b), respectively. The P_{sat} of the



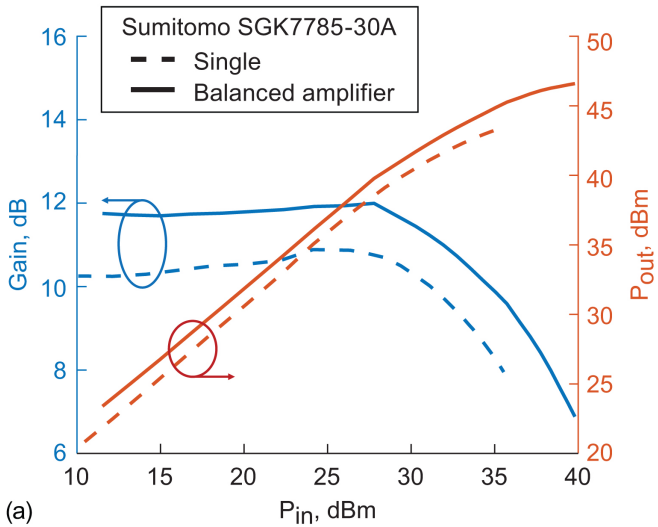
(a)



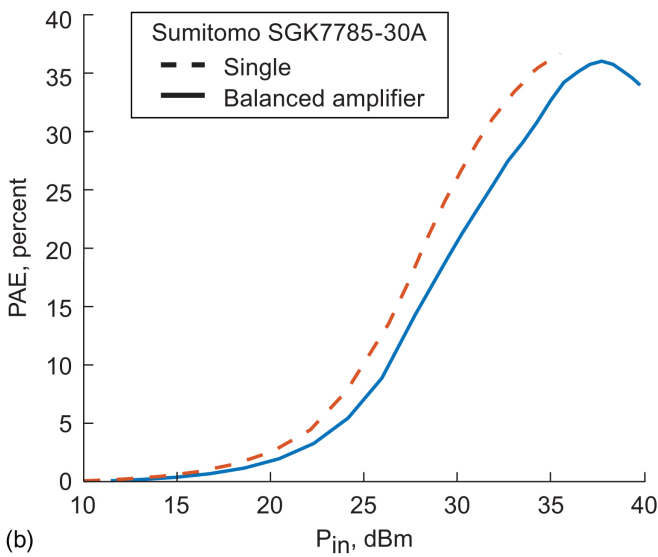
(b)

Figure 10.—(a) Fully packaged Sumitomo SGK7785-30A HPA mounted on a thermoelectric cold plate cooler during characterization, and (b) Balanced amplifier configuration consisting of two Sumitomo SGK7785-30A HPA and Narda Model 4356B hybrid couplers.

single HPA is 43.5 dBm, the corresponding gain is 7.9 dB, and the PAE is 36.7 percent. The P_{sat} of the balanced HPA is 46.7 dBm, the corresponding gain is 6.8 dB, and the PAE is 34 percent. The measured P_{out} and gain and PAE for the balanced HPA as a function of the X-band frequencies at a fixed input drive power and with the thermos electric cooler plate temperature as a parameter are presented in Figure 12(a) and (b), respectively. The P_{out} , gain, and PAE are fairly stable with temperature and frequency.

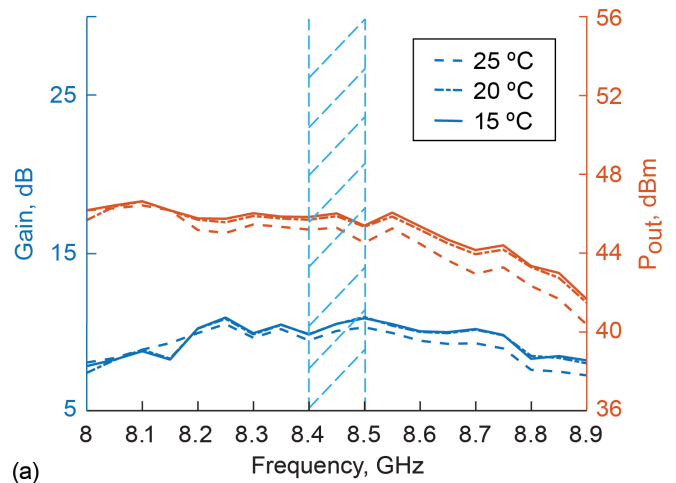


(a)

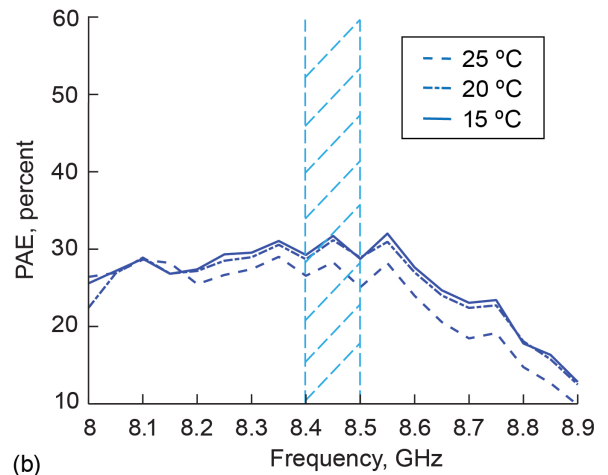


(b)

Figure 11.—(a) Measured output power as a function of the input drive, and (b) Measured PAE as a function of the input drive at X-band (8.475 GHz), for a single HPA Sumitomo SGK7785-30A with $V_D = 24$ V, $V_G = -2.2$ V, $I_{DQ} = 1750$ mA and balanced HPA with $V_D = 24$ V, $V_G = -2.2$ V, $I_{DQ} = 3500$ mA.



(a)



(b)

Figure 12.—(a) Measured output power and gain, and (b) Measured PAE as a function of the X-band frequencies, for the balanced HPA for input drive power in the range of 37.3 to 42.4 dBm and with the thermoelectric cold plate temperature as a parameter.

3.6 GaN Pulsed High-Power Amplifier (HPA)

The pulsed HPA is realized with a Cree Model CGHV96100F2 MMIC, which has GaN HEMT on SiC substrate that are internally matched to 50 ohms. The MMIC is housed inside a metal/ceramic flanged package for optimal thermal performance (Figure 13). The measured characteristics at X-band under pulsed operating conditions is presented in Figure 14(a) and (b). The P_{sat} of the pulsed HPA is 50.86 dBm, the corresponding gain is 9.86 dB, and the PAE is 34.5 percent.

4.0 Proof-of-Concept (POC) Reconfigurable SSMPM and Test Results

A POC bread board version of the fully assembled reconfigurable SSMPM with the diplexer, pre-amplifiers, SPDT switch, MPA, and HPA presented earlier is shown in Figure 15. The measured output power of the end-to-end S-band, X-band (CW), and X-band (Pulsed) chains are presented in Figure 16. The signal pathways are shown in the inset in Figure 16. The results indicate that the S-band CW chain of the SSMPM delivers P_{sat} of 39 dBm (8 W) for TT&C, the X-band CW chain

delivers P_{sat} of 46 dBm (40 W) for telecommunications, delivers P_{sat} of 39 dBm (8 W) for TT&C, the X-band CW chain delivers P_{sat} of 46 dBm (40 W) for telecommunications, and the X-band pulsed chain without the pre-amplifiers delivers P_{sat} of >50 dBm (>100 W) for radar applications.

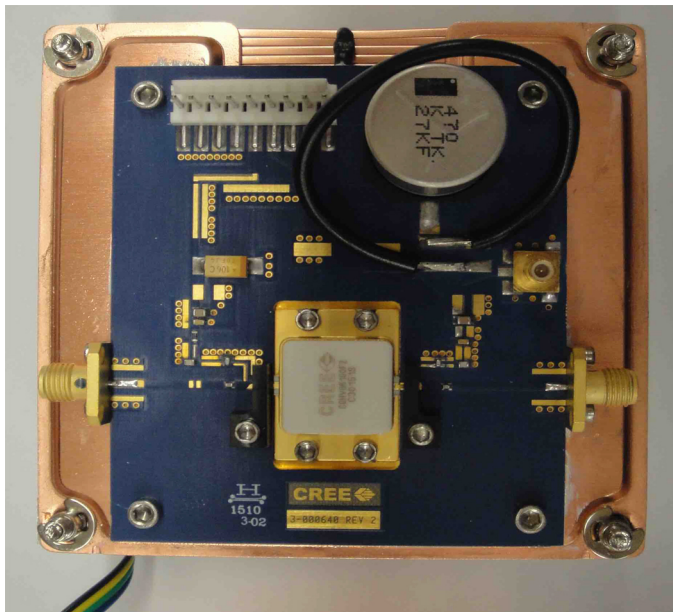


Figure 13.—Cree CGHV96100F2 GaN MMIC based high power pulsed amplifier.

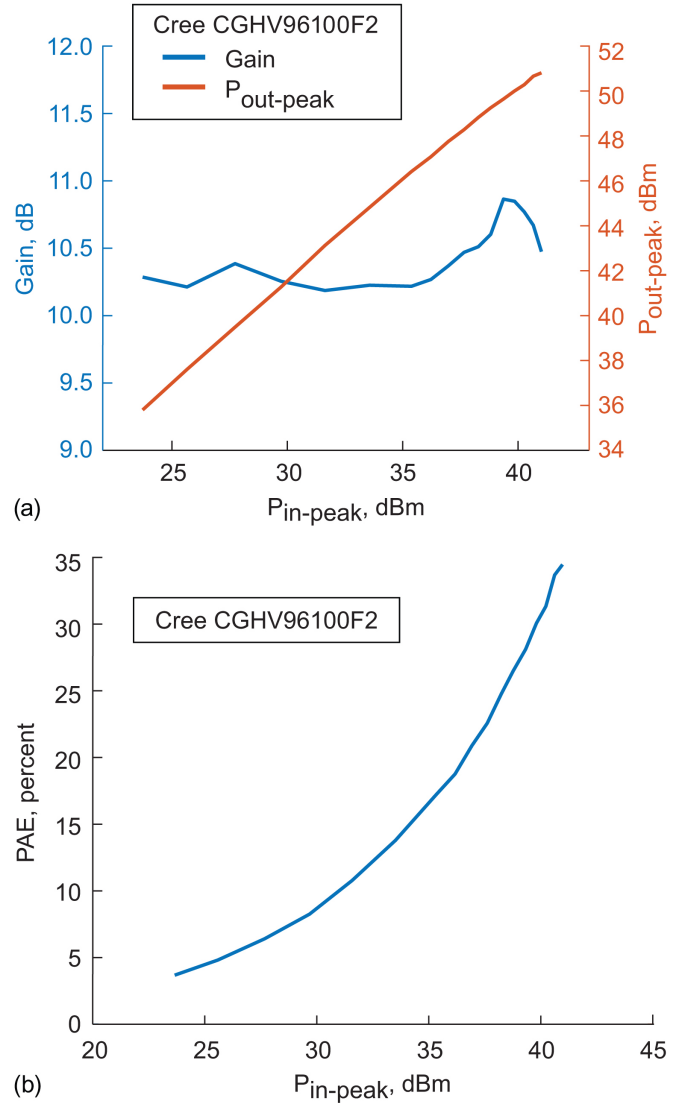


Figure 14.—(a) Measured peak output power and gain as a function of the input drive, and (b) Measured peak PAE as a function of the input drive at X-band (8.475 GHz), for Cree CGHV96100F2 under pulsed conditions, duty cycle = 10 percent, pulse width = 100 μ sec, V_D = 40 V, V_G = -2.2 V, I_{DQ} = 1000 mA.

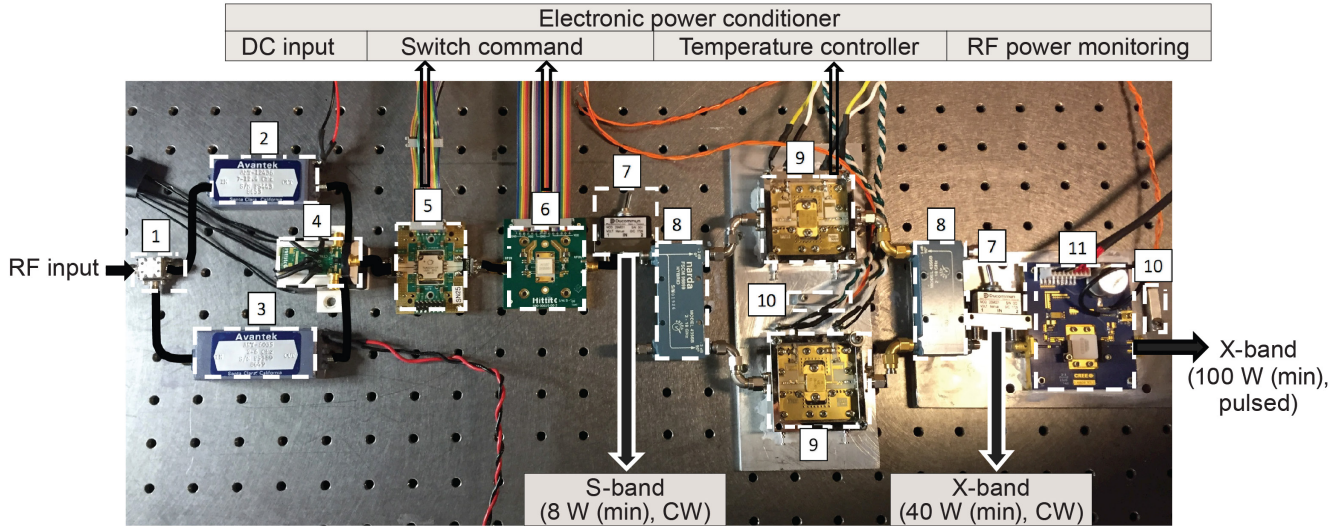


Figure 15.—A POC bread board version of the fully assembled reconfigurable SSMPM as a pre-cursor to a compact SSMPM. (1) Marki Microwaves DPX-4, (2) AvanteK AMT-12436, (3) AvanteK AWT-6035, (4) Analog Devices HMC1118 SPDT, (5) Qorvo TGA2214-CP, (6) Analog Devices HMC1087F10, (7) Ducommun 2SM001, (8) Narda 4356B Hybrid Coupler, (9) Sumitomo SGK7785-30A, (10) Thermocouple, (11) Cree CGHV96100F2.

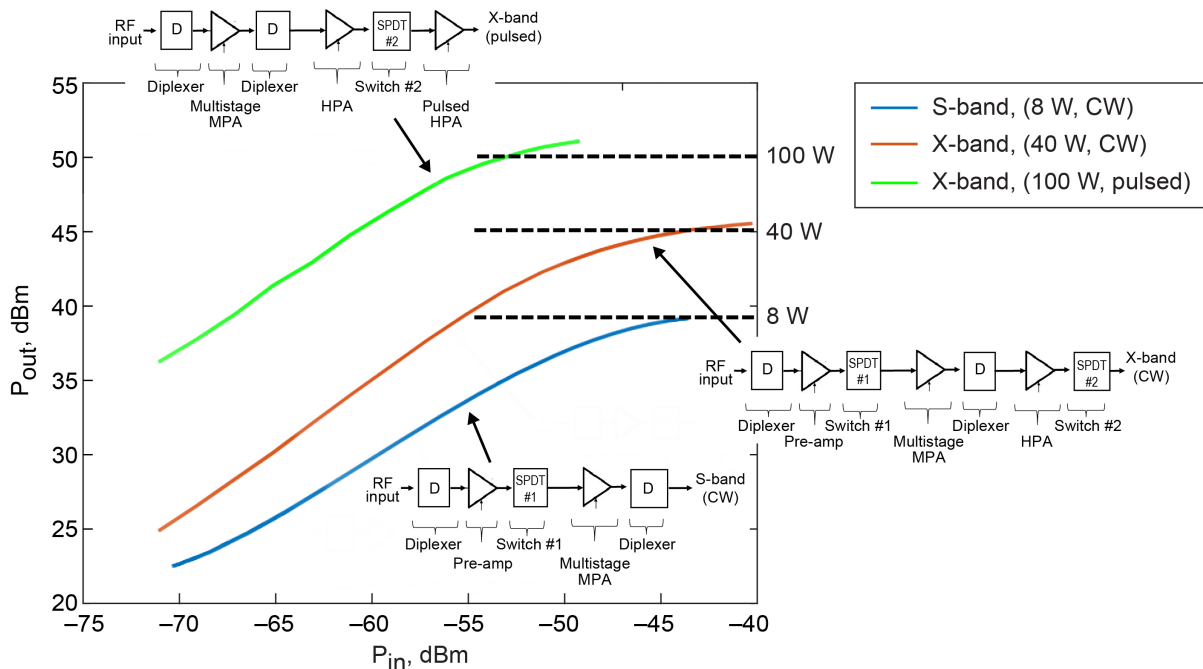


Figure 16.—Measured output power as a function of the input drive for the S-band (CW), X-band (CW), and X-band (Pulsed) signal pathways that are shown in the inset.

5.0 Conclusion and Discussions

The paper presents as a proof-of-concept the design, integration, and performance of a novel reconfigurable, S-/X-band GaN MMIC based, fully SSMPM with a view to miniaturize the overall RF system. The characterization of the individual components as well as the end-to-end performance of each of the S-band and X-band chains of the SSMPM are presented. These results indicate that the S-band CW chain can

deliver P_{sat} of 39 dBm (8 W) for TT&C, the X-band CW chain can deliver P_{sat} of 46 dBm (40 W) for telecommunications, and the X-band pulsed chain without the pre-amplifiers can deliver P_{sat} of >50 dBm (>100 W) for radar applications. In applications that require very high power, waveguide based unequal power divider/combiner can be utilized (Refs. 9 to 14).

Our link budget calculations indicate that the SSMPM with $P_{sat} = 40$ W when coupled to a 10 cm X-band transmit antenna on a low Earth orbiting (900 km) satellite can close a 1 Gbps

(QPSK) data downlink to a 1 m receive antenna on ground with 3 dB margin.

Leveraging upon compound semiconductor devices and novel materials (Ref. 7) will enable the monolithic heterogeneous integration of GaN plus CMOS for realization of a compact SSMPM. The above results indicate that a single SSMPM is capable of being dynamically reconfigured to serve multiple roles such as an amplifier for TT&C, telecommunications, and radar onboard future Earth and planetary exploration spacecrafts.

As a final note, GaN HEMT based X-band power amplifiers when tested under conditions in low Earth orbit during a five-year mission, have shown to be tolerant to a total ionizing dose of 20k rads when placed inside a 2 mm thick aluminum shielding enclosure (Ref. 15).

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