United States Patent

Liu et al.

[54] RECONFIGURABLE OPTICAL INTERCONNECTIONS VIA DYNAMIC COMPUTER-GENERATED HOLOGRAMS

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Filed: Jan. 27, 1994

Int. Cl. [57] G02B 5/32; G02G 1129

U.S. Cl. [57] 359/15; 359/29; 359/320; 365/64

Field of Search [57] 359/9, 11, 15, 359/21, 22, 298, 320; 365/64; 382/31; 395/25

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Abstract

A system for optically providing one-to-many irregular interconnections, and strength-adjustable many-to-many irregular interconnections which may be provided with strengths (weights) \( w_{ij} \) using multiple laser beams which address multiple holograms and means for combining the beams modified by the holograms to form multiple interconnections, such as a cross-bar switching network. The optical means for interconnection is based on entering a series of complex computer-generated holograms on an electrically addressed spatial light modulator for real-time reconfigurations, thus providing flexibility for interconnection networks for large-scale practical use. By employing multiple sources and holograms, the number of interconnection patterns achieved is increased greatly.

3 Claims, 8 Drawing Sheets
OTHER PUBLICATIONS


FIG. 1(a)

FIG. 1(b)

FIG. 2(a)

FIG. 2(b)
FIG. 3(a)

FIG. 3(b)
FIG. 4(a)

FIG. 4(b)
FIG. 5

FIG. 6

APPLIED VOLTAGE (V)

PHASE SHIFT (°)

0 50 100 150 200 250 300 350 400

0 0.5 1 1.5 2 2.5 3 3.5

514.5nm 632.8nm
FIG. 7(a)

FIG. 7(b)
FIG. 8(a)

FIG. 8(b)
FIG. 9(a)

FIG. 9(b)
The invention described herein was made in the performance of work under a NASA contract, and is subject to the provisions of Public Law 96-517 (35 USC 202) in which the contractor has elected not to retain title.

The invention relates to a reconfigurable optical interconnection system for parallel optical computing, neural networks, optical communications and similar applications where reconfigurable electrical interconnection patterns are employed.

Existing electronic interconnection techniques have several physical limitations, such as poor synchronization and low bandwidth, which cannot support the necessary interconnection density (for example, 10^10 processing elements per integrated circuit chip), speed (for example, GaAs based processing elements operating at 100 MHz), and signal bandwidth requirements. To offer a possible solution to these problems, optical interconnection techniques have been suggested. [J. W. Goodman et al., Proc. IEEE, 72, 850 (1984) and A. Husain, SPIE 466, 24, (1984).

Noninterfering free-space optical interconnections possess the potential advantage of massive parallelism, high space bandwidth product (SBWP), high temporal bandwidth product (TBWP), low power consumption, low cross talk and low time skew as compared to electronic interconnections. Free-space optical interconnection techniques have the further potential advantage that they can be used for both chip-to-chip and chip-to-module interconnections. Attention has been given to the use of various optical techniques, such as classical optics, diffraction optics, integrated optics, and holographic optics. For the purpose of implementing some important interconnection architectures, such as perfect shuffle [H. S. Stone, IEEE Trans. Comp., C-20, 153 (1971)], crossbar [M. A. Franklin, IEEE Trans. Comp., C-30, 283 (1981)], multistage interconnection networks (MINs), i.e., crossover networks [J. Jahns et al., Appl. Opt. 27, 3155 (1988)], and Clos networks [C. Clos, Bell Syst. Tech. J., 32, 406 (1953)], and other similar interconnection networks.

Recently, real-time reconfigurable optical interconnection techniques have been developed for parallel optical computing, neural networks, and optical communications using electrically addressed spatial light modulators (ESLM) with computer-generated holograms (CGH). See, for example, A. Marrachi et al., Opt. Lett. 16, 931 (1991); E. C. Tam et al., Technical Digest of the OSA 1991 Annual Meeting, 146 (1991); J. Amako et al., Appl. Opt. 30, 4622, (1991); A. Vanderlugt, SPIE 634, 51 (1986). However, for this ESLM-CGH approach to become practical, higher throughput and more complicated interconnections will be required. For example, prior art implementations represented by U.S. Pat. Nos. 4,946,253; 5,115,497; 5,159,473 and 5,170,269 use a plane wave input beam and a single hologram. The latter patent relies on a deformable mirror device (DMD), which adds complexity to the reconfigurable interconnections but is otherwise also representative of the prior art that relies upon a single plane wave input beam and a single hologram to provide a reconfigurable pattern of interconnections. Such prior-art approaches do not provide sufficient flexibility for a large enough interconnection network with high enough efficiency for large-scale practical use.

An optical system for providing dynamically programable interconnections at spatial points of an array is based on computer-generated holograms (CGHs) applied to an array of pixels of an electrically addressed spatial light modulator (ESLM).

The criteria for computer-generation of the CGHs are array uniformity and diffraction efficiency, where uniformity is measured by a parameter which is defined as the root mean square of the intensities of the focal points in the array and diffraction efficiency is obtained by letting the sum of the powers in the signal spots of the array be divided by the total incident power. An optical system for many spatial interconnections comprises an array of K_xK_y light beams where K_i is an integer greater than one, preferably implemented by a single laser diode and a multiple beam-splitter (MBS), for illuminating through a Fourier transform lens L an array of K_xK_y areas of an ESLM, each area being addressed by a sub-CGH. The combined output from K_xK_y sub-CGHs on the ESLM provides irregular interconnection patterns at an array of light detectors positioned at NxN focal points on the output plane of the ESLM. By effectively placing a K_xK_y binary spatial filter in the front of the ESLM, either as a discrete element or by incorporating it in the sub-CGHs, different irregular interconnection patterns may be provided.

An optical system for many strength-adjustable interconnections comprises a plurality of laser sources 1x(0,1,...,N−1) and a plurality of detectors 1x(s=0,1,...,N−1) interconnected by an ESLM with adjustable strength (weights) W_j, by providing means for extending a linear array of N laser beams in one direction for illuminating each of N different columns of pixels in the ESLM, each of which is addressed by a sub-CGH to generate 1x(s=0,1,...,N−1) beams that illuminate the N detectors positioned in a linear array optically orthogonal to the columns of pixels. Optical means first Fourier transforms the 1x(s=0,1,...,N−1) beams and then combines the N diffraction fields to form N individual beams for N interconnections. The strength of the interconnection sensed at each detector is given by

\[ d_j = \sum_{i=0}^{N-1} W_{ij} \]

where W_{ij} is the weight of the jth element in the interconnection weight matrix determined by the sub-CGHs which are readily adjustable in a computer used to generate the sub-CGHs.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1(a) and 1(b) illustrate the practical pixel spacing of a typical electrically addressed spatial light modulator (ESLM) consisting of m=64 cells of width a and gaps of width b between adjacent cells. FIG. 2(a) illustrates conceptually the prior-art scheme for one-to-many regular reconfigurable optical interconnections, and FIG. 2(b) illustrates an optical implementation for that prior-art scheme where a single computer-generated hologram (CGH) stored in an ESLM is used to generate a regular array of interconnections using a single plane wave input beam.
FIG. 3(a) illustrates conceptually a one-to-many irregular reconfigurable interconnections to be achieved by the present invention for greater flexibility, and FIG. 3(b) illustrates an optical implementation of the irregular optical interconnections of FIG. 3(a).

FIG. 4(a) illustrates conceptually a many-to-many (MxM) irregular reconfigurable optical interconnection network, and FIG. 4(b) illustrates an implementation of the MxM reconfiguration network of FIG. 4(a).

FIG. 5 illustrates conceptually a three-stage crossover network (N=8) utilizing the many-to-many arrangement of FIGS. 3(a) and 3(b) for each stage.

FIG. 6 is a graph of phase shift of a liquid crystal television (LCTV) ESLM versus applied voltage (from 0 to 0.35 V) at two different wavelengths plotted by a dot for 632.8 nm and 514.5 nm.

The parameters are M=64, q=3:1, a=58 pm, T,=0.04, T,=0.96, s=0.04, and $\phi_0=20^\circ$.

FIGS. 7(a) and 7(b) illustrate computer simulation results for an example of one-to-two interconnection CGH using the 1st and 8th order outputs. FIG. 7(a) illustrates phase level distribution, and FIG. 7(b) illustrates diffraction field. The parameters are M=64, q=3:1, a=58 pm, T,=0.04, T,=0.96, s=0.04, and $\phi_0=20^\circ$.

FIGS. 8(a) and 8(b) illustrate computer simulation results for an example of one-to-fourteen interconnection CGH using $\pm 1, \pm 3, \ldots, \pm 13$ order outputs. FIG. 8(a) illustrates phase level distribution, and FIG. 8(b) illustrates diffraction field. The parameters are M=64, q=3:1, a=58 pm, T,=0.04, T,=0.96, s=0.04, and $\phi_0=20^\circ$.

FIGS. 9(a) and 9(b) illustrate computer simulation results for an example of one-to-fifteen interconnection CGH using $\pm 1, \pm 2, \ldots, \pm 15$ order outputs. FIG. 9(a) illustrates phase level distribution, and FIG. 9(b) illustrates diffraction field. The parameters are M=64, q=3:1, a=58 pm, T,=0.04, T,=0.96, s=0.04, and $\phi_0=20^\circ$.

FIGS. 10(a) and 10(b) illustrate experimental results. FIG. 10(a) illustrates results for one-to-fourteen ($\pm 1, \pm 3, \ldots, \pm 13$ order outputs), and FIG. 10(b) illustrates results for one-to-fifteen ($\pm 1, \pm 2, \ldots, \pm 15$ order outputs).

**DETAILED DESCRIPTION OF THE INVENTION**

As shown in FIGS. 1(a) and 1(b), the practical pixel structure of a typical phase electrically addressed spatial light modulator (ESLM) consists of cells of width a and gaps of width b between any two cells. There is also leakage in each cell. In the design of a computer-generated hologram (CGH) which is to be entered into an ESLM, a realistic pixel with leakage should be considered. If m pixels are selected as one period and N subpixels are taken within each pixel to calculate the Fourier spectrum of the profile, the complex amplitude transmittance in the case of one dimension may be expressed as

$$T_i = \frac{1}{m} \left( \sum_{k=1}^{m} a_k \right)$$

where $a_k$ is the amplitude of the kth order output, $a_k$ is the complex amplitude of the kth order output, and $T_i$ is the transmittance of the cell and gap, respectively, and $\phi_k$ is the phase shift corresponding to L different phase levels. The phase shift of the gap and dark cell ($\phi_k=1$) are assumed to be zero and $\phi_k$, respectively. In the implementation of the present invention, it is preferable to so prepare the CGHs that the transmittances of all cells are at the same T, level, and only the phase shift is modulated.

The fast Fourier transform (FFT) algorithm is used to calculate the Fourier spectrum of Eq. (1), and annealing of the Fourier spectrum is achieved, such as by the simulated annealing iteration algorithm described by S. Kirkpatrick et al., Science, 220, 671 (1983) incorporated herein by reference to obtain the desired diffraction patterns. There are two criteria for designing the CGH: (1) array uniformity and (2) diffraction efficiency [M. R. Feldman et al., Opt. Lett., 14, 479 (1989)]. The uniformity of the array is measured by a parameter which is defined as the root mean square error of the intensities of the focal points in the array. The diffraction efficiency is obtained by letting the sum of the powers in the signal spots of the array be divided by the total incident power.

An overall error mass function $E$ is defined as follows:

$$E = E_1 + E_2$$

where $I_2$ is the intensity of the kth diffraction order, $I_2$ is the mean value averaged over all the diffraction orders, $K$ is the total number of the focal points in the array, and $C_1$ and $C_2$ are the optimization coefficients between zero and one. The desired array of spots may be designed based on an error tolerance (or a minimum tolerable error mass function) given a priori.

A two-dimensional CGH may be obtained by the superposition of two one-dimensional CGH structures. However, for increasing diffraction efficiency, a two-dimensional simulated annealing iteration algorithm may be used to search directly for an optimal two-dimensional CGH structure.

**One-to-Many Regular Interconnections**

Reconfigurable one-to-many regular interconnections illustrated conceptually in prior-art FIG. 2(a) comprises a single laser light source 10 and an array (N x N, N=3) of detectors 11 that can be implemented directly by means of a single complex amplitude (or phase) computer-generated hologram (CGH) on an electrically addressed spatial light modulator (ESLM) which is illuminated by a laser light source S, e.g., a laser diode, through a Fourier transform lens L, as shown in FIG. 2(b). N x N multiple focal points will be formed on detectors 11 positioned on the focal plane of the ESLM. The detectors illuminated by the light from the source transmitted through the ESLM under control of the CGH conduct and thus provide interconnections between the two terminals of the detectors (photodiodes) at the spatial points of the array of detectors. The interval between two neighboring focal points can be controlled by changing the system parameters, such as object distance $d_o$, focal length of the lens and the separation q between lens L and the ESLM.

**One-to-Many Irregular Interconnections**

Reconfigurable one-to-many irregular interconnections, illustrated conceptually in FIG. 3(a) as again comprising a
light source 10 and an array of detectors 11, may be implemented by providing a source of multiple beams, such as a fixed K1 x K1 (K1=2) multiple beam splitter (MBS) 14 shown in FIG. 3(b) to provide K1 x K1 parallel channels that illuminate K1 x K1 different CGHs from a digital computer 12 on the spatial light modulator 13. Each channel provides a pattern of regular interconnections illustrated as being one pattern in each of four quadrants for the case of K1=2, but the combination of these K1 x K1 channels realize multiple irregular interconnections at an array of detectors 11. If a K1 x K2 binary spatial filter array is placed in front or in back of the spatial light modulator 13, thereby producing a pattern of spots of no light onto the detectors from the spatial light modulator, whereby 2K1 x 2K1 different irregular interconnection patterns (equal 16 for the case of K1=2) may be constructed by controlling the individual transmittance (opaque or transparent) in the filter array. The function of such a binary spatial filter 15 may be implemented in the digital computer by modifying the CGH generated according to the effect of the binary spatial filter superimposed on the CGH.

Strength-Adjustable Many-to-Many Interconnections

Strength-adjustable many-to-many interconnections, illustrated conceptually in FIG. 4(a) may be implemented as shown in FIG. 4(b). Referring first to FIG. 4(a), a plurality of sources sji (i=0,1, . . . , N-1) and N detectors dji (j=0,1, . . . , N-1) are interconnected with adjustable weights wji. Referring now to FIG. 4(b), assume the same linear array arrangement of sources 20 and detectors 21, but one oriented vertically and the other horizontally, although optical elements could be used to rotate the linear array of detectors 90° so they are oriented parallel to the linear array of the sources. The laser beams from the sources 20 are extended and collimated in the vertical direction by means of a spherical lens L1 and a cylindrical lens CL1, to illuminate N different columns of an electrically addressed spatial light modulator (ESLM) 22. Each column of the spatial light modulator is used to store a sub-CGH used to generate 1xj (j=0,1, . . . , N-1) beams. A set of spherical lenses L2 set and cylindrical lens CL2 is used to first Fourier transform the N sub-CGHs and then combine these N diffraction fields so that these N different sub-CGHs may form N individual beams for N interconnections. The power sensed at an output detector dji may be written as

\[ d_j^{(i)} = \sum_{j=0}^{N-1} w_{ji} \]

where wji is the jth element in the interconnection weight matrix (IWM) determined by the sub-CGHs which are, of course, adjustable in the computer 12 used to generate the sub-CGHs. In that manner, weighted interconnections between the two terminals of the detectors are provided at the spatial points where the detectors are positioned. A variety of networks can be realized by the strength-adjustable many-to-many interconnection technique. Networks such as perfect shuffler and crossbar may be implemented via a single layer strength-adjustable many-to-many interconnection. The multistage interconnection networks such as crossover [J. Jahns et al., Appl. Opt., 27, 3155 (1988)], Clos [C. Clos, Bell Sys. Tech. J., 32, 406 (1953)], Benes [V. Benes, Bell Sys. Tech. J., 41, 1481 (1962)], Omega [D. H. Lawire, IEEE Trans. Comput., C-24, 1145 (1975)] and Baseline [C. Wu et al., IEEE Trans. Comput., C-29, 694 (1980)] may be implemented by cascading several strength-adjustable many-to-many interconnections. As an example, a three-stage crossover interconnection shown in FIG. 5 may be implemented as follows. The jth output permutation of the power of the crossover may be represented as

\[ d_j^{(3)} = \begin{cases} d_j^{(1)} + d_j^{(1)} d_j^{(2)}, & 0 \leq j < N/2 - 1 \\ d_j^{(1)} + d_j^{(2)} d_j^{(1)}, & N/2 \leq j < N - 1 \end{cases} \]

For the third stage, the permutation of the output dji may be written as

\[ d_j^{(3)} = \begin{cases} d_j^{(1)} + d_j^{(1)} d_j^{(2)}, & 0 \leq j < N/4 - 1 \\ d_j^{(1)} + d_j^{(2)} d_j^{(1)}, & N/4 \leq j < N/2 - 1 \\ d_j^{(2)} + d_j^{(2)} d_j^{(2)}, & N/2 \leq j < 3N/4 - 1 \\ d_j^{(3)} + d_j^{(3)} d_j^{(3)}, & 3N/4 \leq j < N - 1 \end{cases} \]

For N=8, the first stage IWM may be written as

\[ w^{(1)} = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix} \]

For the second stage,

\[ w^{(2)} = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \end{bmatrix} \]

and for the third stage

\[ w^{(3)} = \begin{bmatrix} 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix} \]

From these results, eight 1x2 strength-adjustable sub-CGHs of SBWP=8 are needed for each stage.

Computer Simulations and Experimental Results

As noted with reference to FIGS. 1(a), (b) and (c) and 2(a), (b), the computer-generated holograms may be amplitude or phase, although amplitude CGHs are implied
throughout. The following computer simulations and results were carried out with phase holograms for convenience in implementation of the computer program since they are equivalent to a Silicon-tendon liquid-crystal television (LCTV) ESLM operated in phase mode was available. First it was necessary to establish a look-up table between the phase shift on the spatial light modulator ESLM and the applied voltage for phase control. FIG. 6 gives the experimental result of the relationship between phase shift and applied voltage. The results of using a He-Ne laser and an Ar+ laser are shown. The dark phase \( \phi_0 \) was not considered in the figure. The phase shift is measured by means of a Mach-Zehnder interferometer. Other parameters of the LCTV are measured and the results are \( q=3:1, a=58 \mu m, T_0=0.04 \) and \( s=4\% \). \( T_N \) is varied with an unexpected amplitude \( \pm 10\% \) within the region from 0–3.5 V because of the imperfect alignment of the liquid crystal molecules on the thin film transistor substrate [I. Marrakechi et al., (1991), supra]. An average value, \( T_N=0.96 \) is used in the computer simulations for simplifying the calculation process. The results of (a) pixel structures and (b) relative intensities for the computer simulation of several kinds of CGHs useful for making different interconnections are presented in FIG. 7 through FIG. 9. FIGS. 7(a) and 7(b) illustrate computer simulation results for an example of one-to-two interconnection CGH using the 1st and 8th order outputs. FIG. 7(c) illustrates phase level distribution, and FIG. 7(d) illustrates diffraction field. The parameters are \( M=64, q=3:1, a=58 \mu m, T_0=0.04, T_N=0.96, s=0.04 \), and \( \phi_0=20^\circ \). FIGS. 8(a) and 8(b) illustrate computer simulation results for an example of one-to-fourteen interconnection CGH using \( 1\leq m \leq 13 \) order outputs. FIG. 8(a) illustrates phase level distribution, and FIG. 8(b) illustrates diffraction field. The parameters are \( M=64, q=3:1, a=58 \mu m, T_0=0.04, T_N=0.96, s=0.04 \), and \( \phi_0=20^\circ \). FIGS. 9(a) and 9(b) illustrate computer simulation results for an example of one-to-fifteen interconnection CGH using \( 1\leq m \leq 15 \) order outputs. FIG. 9(a) illustrates phase level distribution, and FIG. 9(b) illustrates diffraction field. The parameters are \( M=64, q=3:1, a=58 \mu m, T_0=0.04, T_N=0.96, s=0.04 \), and \( \phi_0=20^\circ \). The experimental results demonstrating \( 1\times 14 \) and \( 1\times 15 \) interconnections are given in FIG. 10(a) and 10(b), which illustrate experimental results. FIG. 10(a) illustrates results for one-to-fourteen (\( 1\leq m \leq 13 \) order outputs), and FIG. 10(b) illustrates results for one-to-fifteen (\( 1\leq m \leq 15 \) order outputs). The consideration described above is based upon the assumption that \( N \) inputs are incoherent. However, if the input is a coherent source instead of the incoherent source array, constructive and destructive interference may affect the output. The power change caused by the interference on the vertical axis is not very significant if the ratio \( R \) between the horizontal dimension of the sub-CGH and the interval between the two nearest neighbor sub-CGHs is much less than one. For the same LCTV-ESLM, \( M=64, L=8, N=8, \) and \( R=0.1 \), computer simulation results show that the percentage decrease of the power is less than 5%.

We claim:

1. A system for providing dynamically programable optical interconnections comprising

   a spatial light modulator having a \( K_x \times K_y \) array of areas of transmittance pixels, said pixels of each area being adapted to separately modify transmittance of light through each pixel thereof in response to a plurality of separate subholograms defined by a pattern of electrical signals, one subhologram for each area of said \( K_x \times K_y \) array,

   means for producing a plurality of parallel light beams, including a single source of light and a multiple beam splitter for producing said plurality of parallel light beams, one light beam for each of a plurality of areas of said spatial light modulator, each area having a separate array of pixels, each light beam illuminating for transmittance a separate one of said \( K_x \times K_y \) areas of said spatial light modulator, and

   an array of light detectors positioned to receive said light beams transmitted through said \( K_x \times K_y \) areas of said spatial light modulator, whereby light detected at each of said light detectors provides conduction in response to light received, thereby providing many irregular interconnections in response to a plurality of light beams in patterns determined by said plurality of subholograms,

   a computer for generating said plurality of subholograms for producing rapid reconfigurable interconnections at said array of light detectors, and

   a binary spatial light filter positioned in the path of light through said spatial light modulator to impose a pattern of pixel spots of no light transmitted through said spatial light modulator, thereby increasing the number of patterns of irregular interconnections producible at said array of detectors.

2. A system for providing dynamically programmable optical interconnections comprising

   a spatial light modulator having a \( K_x \times K_y \) array of areas of transmittance pixels, said pixels of each area being adapted to separately modify transmittance of light through each pixel thereof in response to a plurality of separate subholograms defined by a pattern of electrical signals, one subhologram for each area of said \( K_x \times K_y \) array,

   means for producing a plurality of parallel light beams, each light beam illuminating for transmittance a separate one of said \( K_x \times K_y \) areas of said spatial light modulator, and

   an array of light detectors positioned to receive said light beams transmitted through said \( K_x \times K_y \) areas of said spatial light modulator, whereby light detected at each of said light detectors provides conduction in response to light received, thereby providing many irregular interconnections in response to a plurality of light beams in patterns determined by said plurality of subholograms,

   a computer for generating said plurality of subholograms for producing rapid reconfigurable interconnections at said array of light detectors,

   a plurality of different columns of cells in said spatial light modulator, each column of cells receiving a separate computer generated subhologram, and including

   a plurality of sources of light, \( s_p \),

   means for extending and collimating in one direction each of said sources of light to illuminate corresponding ones of said plurality of columns of pixels in said spatial light modulator,

   means for recombining light transmitted by said plurality of columns of pixels to produce a plurality of individual beams, one beam for each row, and

   said array of detectors comprises a column of detectors, \( d_j \), one for each of said individual beams, thereby producing from each detector an interconnection, the strength of conduction of each interconnection of each \( N \) detectors being equal to
whereby each column of cells in said spatial light modulator receives a separate computer generated subhologram to generate $[i]=0,1, \ldots, N-1$ beams transmitted to $[i]=0,1, \ldots, N-1$ detectors in response to $[i]=0,1, \ldots, N-1$ sources of light and $w_{ji}$ is the $j$th element of an interconnection weight matrix determined by said computer generated subholograms, thereby providing weighted interconnections at said detectors.

3. A system for providing dynamically programmable optical interconnections comprising

a spatial light modulator having a $K_1 \times K_1$ array of areas of transmittance pixels, said pixels of each area being adapted to separately modify transmittance of light through each pixel thereof in response to a plurality of separate subholograms defined by a pattern of electrical signals, one subhologram for each area of said $K_1 \times K_1$ array,

means for producing a plurality of parallel light beams, each light beam illuminating for transmittance a separate one of said $K_1 \times K_1$ areas of said spatial light modulator, and

an array of light detectors positioned to receive said light beams transmitted through said $K_1 \times K_1$ areas of said spatial light modulator, whereby light detected at each of said light detectors provides conduction in response to light received, thereby providing many irregular interconnections in response to a plurality of light beams in patterns determined by said plurality of subholograms, including a plurality of cascaded spatial light modulators between said light sources and said detectors for providing multistage crossover interconnections, wherein the light beams of each spatial light modulator become the sources of light beams for the next spatial light modulator in cascade, and the $j$th output permutation of the power of the crossover is represented by

$$d_j^{(m)} = \sum_{i=0}^{N-1} w_{ji},$$

for the first stage where the stage number is denoted by the superscript in parenthesis, and $N$ must be an integral power of two, wherein the output of the first stage is used as input to the second stage the permutation of the second stage output $d_2^{(m)}$ may be written as

$$d_2^{(m)} = \begin{cases} d_1^{(m)} + d_1^{(m)}, & 0 \leq j \leq N/2 - 1 \\ d_1^{(m)} + d_1^{(2m)}, & N/2 \leq j \leq N - 1 \end{cases}$$

and so forth to the permutation of the last stage.

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