ABSTRACT

This paper presents the design, fabrication and characterization of a wireless capacitive pressure sensor with directional RF chip antenna that is envisioned for the health monitoring of aircraft engines operating in harsh environments. The sensing system is characterized from room temperature (25°C) to 300°C for a pressure range from 0 to 100 psi. The wireless pressure system consists of a Clapp-type oscillator design with a capacitive MEMS pressure sensor located in the LC-tank circuit of the oscillator. Therefore, as the pressure of the aircraft engine changes, so does the output resonant frequency of the sensing system. A chip antenna is integrated to transmit the system output to a receive antenna 10 m away.

The design frequency of the wireless pressure sensor is 127 MHz and a 2% increase in resonant frequency over the temperature range of 25 to 300 °C from 0 to 100 psi is observed. The phase noise is less than -30 dBc/Hz at the 1 kHz offset and decreases to less than -80 dBc/Hz at 10 kHz over the entire temperature range. The RF radiation patterns for two cuts of the wireless system have been measured and show that the system is highly directional and the MEMS pressure sensor is extremely linear from 0 to 100 psi.

KEYWORDS

Wireless capacitive pressure sensor system, MEMS, chip antenna, passive and active components, high temperature applications, health monitoring.

INTRODUCTION

Sensing systems for health monitoring in harsh environments, specifically high temperatures, are of extreme importance and are highly sought for many applications from the oil and gas industry to monitoring the performance of aerospace systems. Currently, the largest user of high-temperature sensing electronics is the downhole oil and gas industry. In the past, drilling operations were at maximum temperatures of 150°C to 175°C, but declining reserves of easily accessible natural resources, coupled with the advances in technology, have motivated the industry to drill deeper, and as a result, the electronics must operate as high as 300°C [1]. The automotive industry is another field where high temperature sensor electronics are widely used. The need to locate sensors, signal conditioning and control electronics closer to the heat sources will be an essential requirement [2]. For example, cylinder pressure temperatures are monitored up to 300°C, and exhaust sensing can be as high as 850°C [3]. Additionally, in the aerospace industry, aircraft engines that use intelligent controls and health monitoring require a variety of sensors [4]. Many of the sensors operate at temperatures at and above 500°C, requiring electronics that can function at these temperatures to assure essential health monitoring [5, 6]. Combining health monitoring in harsh environments with real-time wireless transmission of diagnostic data to a remote receiver greatly enhances the sensing system functionality and importance [7]. Although efforts in this area have been made, they often used large thin film spiral inductors as the radiating element [8, 9]. However, large inductors are very sensitive to changes in environmental conditions, including a shift in the output frequency if the inductor is placed near a metallic surface. In addition, the large spiral inductor greatly increases the overall size of the sensing system.

In this paper, the development of a wireless capacitive MEMS pressure sensor with integrated directional chip antenna designed for high temperature applications is described. To the best of our knowledge, this is the first effort to integrate a chip antenna with a sensing system for harsh environment applications. The directional antenna allows for maximum power transfer to the receiver. A matching network was added to the input of the chip antenna to shift its resonant frequency to that of the oscillator/pressure sensor. The wireless sensor system on chip performance as a function of temperature is demonstrated from room temperature (25°C) to 300°C for a pressure range of 0 to 100 psi. Additionally, two cuts of the radiation pattern, vertical and horizontal, of the wireless system are presented and show highly directional capabilities.

PRESSURE SENSOR DESIGN

The capacitive micro-electromechanical systems (MEMS) pressure sensor was developed by Sporian Microsystems and is shown in Fig. 1. One electrode of the capacitive sensor is fabricated on the deflecting chamber membrane which forms a sealed cavity and the second electrode is on the fixed silicon carbide nitride (SiCN) substrate. The sealed cavity is flip-chip bonded onto a SiCN substrate with gold (Au) contacts so the tow electrodes form a parallel plate capacitor. The Sporian Microsystem pressure sensor was characterized with an Agilent B1500A Semiconductor...
Device Analyzer (SDA) and a high temperature/pressure chamber (HTPC), shown in Fig. 2. The SDA was calibrated to the leads of the pressure sensor inside the HTPC at room temperature to remove the effects of the chamber and cabling. The pressure sensor capacitance was measured at 1 MHz from 0 to 100 psi at 25°C, and the results are shown in Fig. 3. The pressure sensor was measured twice to demonstrate reliability. The capacitance is 5.15 pF at 0 psi and decreases to 4.41 pF at 100 psi, resulting in a ΔC of 0.74 pF.

Figure 1: Sporian Microsystem capacitive pressure sensor.

Figure 2. High temperature pressure chamber.

Figure 3: Capacitance versus pressure at 25°C.

**OSCILLATOR WITH PRESSURE SENSOR**

A circuit schematic of the Clapp-type oscillator with an integrated capacitive pressure sensor is shown Fig. 4. The resonant frequency of the oscillator greatly depends on the values of the inductor, $L_T$, and the pressure sensor, $C_T$. As the pressure increases, the capacitance of the pressure sensor decreases, causing the resonate frequency of the sensing system to increase. More information on the design methodology for this oscillator can be found in [10].

The sensing system was fabricated on an alumina CoorsTek Superstrate 996 with a dielectric constant of 9.9 and a substrate thickness of 500 μm. The metallization consists of a Ti/Au (0.15/1.5 μm) layer, which was deposited using an E-beam evaporator. Finally, gold wire bonds were used to make electrical connections between components and pads.

A Cree CRF24010D SiC MESFET is the active device used in this system. The values for $L_T$, $C_1$, and $C_2$ are 390 nH, 13 pF, and 40 pF, respectively. $L_T$ is a Johanson RF wire wound 390 nH chip inductor. The inductor was tested from room temperature to 400°C and showed less than a 2% decrease in inductance up to 350°C, after which the inductance dropped by 8% at 400°C. It should be noted that the inductors were left to cool down to room temperature and re-measured the following day and found to retain their original inductance value of 390 nH, which indicates that the inductors were not permanently modified by the high temperature measurements. The chip capacitors, $C_1$ and $C_2$, from Compex type C-50, consist of a thin titanate dielectric material, with dielectric constant of 40 and thickness of 76.4 μm with gold metallization on both sides. $C_1$ is a single chip capacitor with a value 13 pF and $C_2$ is a single chip capacitor of 40 pF. Similar to the chip
inductors, the Compex chip capacitors were also characterized separately up to 500°C and show less than 1% variation in capacitance up 400°C, after which the capacitance falls off by 5% at 500°C. C_T is the Sporian pressure sensor that was described previously. L_d, C_d, and R_g are the DC bias circuits and are off chip and are left at room temperature. Their values are 2.5 mH, 1.5 µF, and 5 kΩ, respectively. The active pressure sensor system was designed to operate at 127 MHz and increase to 130 MHz at 0 psi and 100 psi, respectively, at room temperature, based on the change in capacitance of the Sporian pressure sensor. Although the design frequency was chosen based on currently available chip capacitors and inductors, it can easily be modified for other design frequencies.

CHIP ANTENNA

A Johanson Technology ISM chip antenna (P/N: 0169AT62A0010) was used to transmit the RF output signal from the oscillator with integrated pressure sensor to a receive antenna 10 m away. The nominal resonant frequency of the chip antenna is 169 MHz when using Johanson’s suggested circuit board and matching network at the input of the antenna. However, the resonant frequency of the antenna is highly dependent on several factors, such as antenna placement with respect to a ground plane (if used), matching network, and substrate/circuit board material. In this case, the antenna is epoxied to an alumina substrate with no ground plane. The alumina substrate is chosen due to its ability to withstand high temperatures [11, 12]. The chip antenna with coplanar waveguide (CPW) feed line and matching network is shown in Fig. 5. The return loss (S11) of the chip antenna with and without a matching network, shown in Fig. 6, was characterized with the Agilent E8361C Precision Network Analyzer (PNA) and ground-signal-ground (GSG) probes to facilitate connection to the CPW feed lines.

![Figure 5. Chip antenna, matching network and CPW feed line.](image)

![Figure 6. Return Loss of chip antenna with and without matching networks.](image)

Since the return loss of the chip antenna has a resonant frequency of 160 MHz, a matching network must be added. The matching network designed for this application must decrease the resonant frequency to that of the original design frequency as well as increase the operational bandwidth to accommodate the change in oscillator/pressure sensor frequency change due to the change in pressure. The matching network consisted of two Johanson 390 nH wire wound chip inductors placed in shunt at the input to the antenna, which results in an equivalent inductance of 195 nH. Accordingly, the reflection coefficient of the chip antenna is at least -5 dB over the operational bandwidth of 125 MHz to 145 MHz.

![Figure 4. Circuit schematic of oscillator with integrated capacitive pressure sensor.](image)
WIRELESS PRESSURE SENSOR

The chip antenna and matching network were added to the output of the oscillator/pressure sensor via gold wire bonds, as seen in Fig. 7. All components were mounted onto alumina substrates using high temperature silver epoxy from Cotronics. The pressure sensor is 15 mm x 5 mm, the oscillator is 7 mm x 7 mm, and the chip inductor is 25 mm x 5 mm, which, if optimized, can be realized in a footprint of approximately 250 mm². When compared to [8], with an approximate area of 2,580 mm², the percentage of area reduction is nearly 90%. For a room temperature, unpressurized test of the system, the sensor was mounted on a styrofoam block and placed on a rotary stage. The test was performed in a typical laboratory environment with lab benches and cabinets that acted as reflectors. The signal was transmitted to a Pixel broadband loop antenna over a distance of 10 m. A horizontal and vertical cut of the radiation patterns for the wireless pressure sensing system are presented in Fig. 8 and show the system to be directional with a maximum power of -23 dBm at 25°C.

The high temperature/pressure chamber, shown in Fig. 2, developed at NASA Glenn Research Center, was used to heat and pressurize the wireless pressure sensor up to 300°C and 100 psi, respectively. The chamber was fitted with feed-throughs so biasing can be applied and the temperature can be monitored at several different locations throughout the chamber. Because the pressure chamber is metal with a small glass window, far field antenna measurements could not be performed. Instead, a dipole antenna was used as the receive antenna 1 m above the glass window.

Figure 7. Compact wireless pressure sensor with directional antenna.

The wireless pressure sensing system was characterized from 25 to 300°C for a pressure range of 0 to 100 psi, in steps of 10 psi, with an Agilent E4440A Series Spectrum Analyzer serving as the receiver. The results are shown in Fig. 9. The output frequency with respect to temperature, over the entire pressure range is linear. The sensitivity of the sensor (Δf/ΔP MHz/psi) is essentially independent of temperature, with values of 2.5, 2.6, 2.8 and 2.8 % for 25, 100, 200, and 300°C, respectively.

Figure 8. Horizontal and vertical radiation patterns of wireless pressure sensor.

Figure 9. Wireless pressure sensor frequency vs. pressure at 25, 100, 200, and 300°C.
The power output vs. frequency was recorded at 300°C for pressures of 0, 50 and 100 psi and is shown in Fig. 10. The power increases slightly and shifts by approximately 2.5 MHz as a function of pressure. The phase noise of the wireless pressure sensor is shown in Fig. 11 at 0 psi for temperatures of 25, 100, 200, and 300°C. The data show that the 1 kHz offset is below -30 dBc/Hz and the 10 kHz offset is below -80 dBc/Hz. The phase noise slope is -20 dBc/decade above 10 kHz. The two spurs evident at 2 kHz are likely due to the heater power supply fluctuating to maintain temperature.

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**REFERENCES**


